# BIG MULTIPORT MEMORY SYSTEM

	REVISION RECORD
Revision	Notes
Aug. 1978	Original Printing
	N

BIG MULTIPORT MEMORY SYSTEM ND-06.007.01



NORSK DATA A.S.

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# 1 BIG MULTIPORT MEMORY SYSTEM

### 1.1 GENERAL

The Big Multiport Memory System (referred to as BMPM) is a high speed flexible and modular memory system used for two major applications:

- Increased memory band width to enable for higher total data transfer rates to and from memory.
- Allow for multiprocessor and/or multi-device communication via a common memory system.

This is accomplished by using a number of independent memory banks accessed by a number of independent memory ports. The system is modular with respect to banks, channels and storage capacity.

Each memory channel has an address range of 2048K words and may be connected from 1 to 10 ports.

Single bit error correction and multiple error detection is standard.

A special service channel is provided for a number of operations and maintenance purposes.

### 1.2 THE MEMORY SYSTEM

A memory system (refer to Figure 1.1) is a collection of independent *banks*, each covering a definite address space. Every bank is linked together and to the outside world by *Channels*. The entrance for channels is called *Ports*. The Channels in the system are driver from *Sources*.

The Sources are interfaces between the memory system and the unit demanding access. In order to maintain a well-defined intersection the sources are not considered part of the memory system.

In addition to the ordinary channels a *special service channel* exists. All banks are linked to the service channel which is driven from an I/O interface.

- 1.3 SYSTEM PARTS
- 1.3.1 Crate

For the following discussion refer to Figure 1.2.

Although the bank is logically the basic building block, the card rate is physically the basic unit, containing one or two banks, an X-bank and a Y-bank. This is done purely for space and wiring reasons. In BMPM there is no logic in common between the two banks as it is on the former model. Yet the service channel has common logic since it will always be connected in a predetermined manner.

A maximum of 8 crates can be used in a Memory system.

Figure 1.1: Multiport Configuration — Example

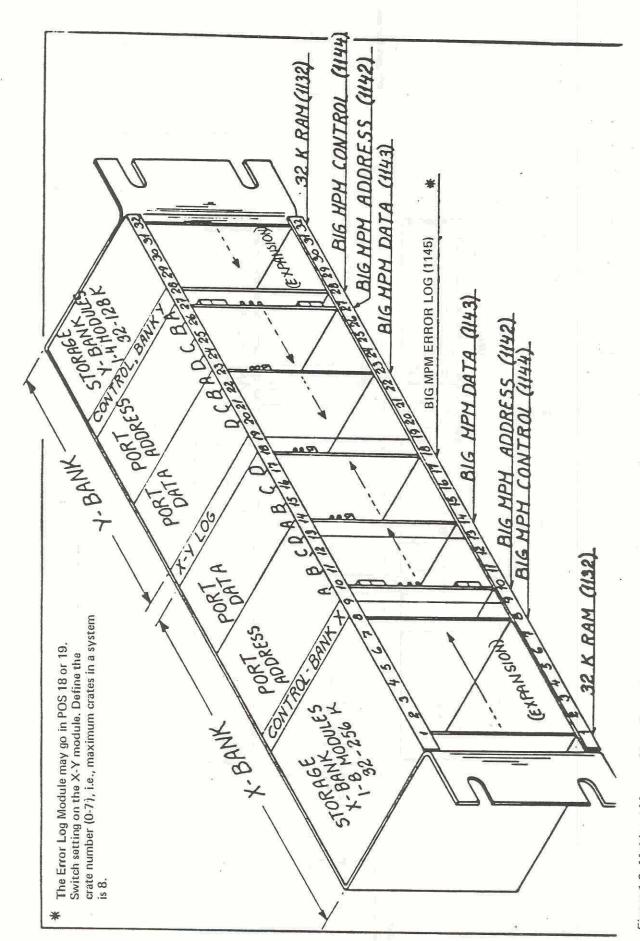


Figure 1.2: Multiport Memory (1 card crate)

### 1.3.2 Bank

A bank is an independent memory unit that can cycle by itself. It also has it's own timing and control logic.

Two banks can be located in one card crate, one X-bank and one Y-bank.

A complete bank consists of the following parts:

- \* Controller (1144)
- \* Storage (1132's)
- \* Ports (1142, 1143)
- \* Error logic (1145)

# 1.3.3 *Controller* (1144)

It is essential that a bank has a private controller which allows it to operate independent of other banks.

The controller solves two major tasks.

- To serve as a "switchboard" between the ports on one side and the storage on the other
- 2. To refresh the MOS memory elements at regular intervals.

Since up to 4 ports (A - D) may access the same storage, a priority network allocates the storage to one port for one storage cycle at a time.

The ports and refresh are assigned a fixed priority which is, beginning at the highest priority:

New request from same port as previous one, REFRESH, PORT A, PORT B, PORT C, PORT D

When two ports make requests simultaneously the highest priority port will be served first.

Note! It is not possible for one high priority port to lock out the lower priority ones. The reason is that the request must toggle on and off.

Even if a channel is capable of absorbing everything it could get from memory, it must withdraw the request for a moment to prepare for the next. At that very moment the lower priority port will be granted access.

However, two high priority ports are able to lock the bank for lower priority ports.

Periodic refresh is necessary in MOS memories to avoid decline of information. This is accomplished by regularly accessing all memory cells. These accesses (one each 15  $\nu$ s) may be treated similar to accesses from the ports.

The refresh request circuitry and refresh address counter is located in the controller (1144).

# 1.3.4 Storage (1132's)

An X-bank has room for 8 storage modules and a Y-bank has room for 4 storage modules. Each module contains 32K x 21 memory cells (bits). Hence, one X-bank is sufficient to cover all NORD-10/S' primary storage requirements of 256K words.

Internally in the bank the address range is always from 0 in 32K increments up to 256K. Hence, an address transformation is normally required at each port. (Refer to Section 1.4 regarding addressing.)

The 18 address bits required for a 256K address range are used as shown in Figure 1.3.

The memory integrated circuits (MIC) contain 16K x 1 bit. Hence, two MIC's are needed per data bit on a 32K module. Since storage of 21 data bits are required each storage module contains 42 MIC's. The 5 extra bits are used by the control code attached to the 16 data bits. (Error checking and correction is described in Section 1.5.)

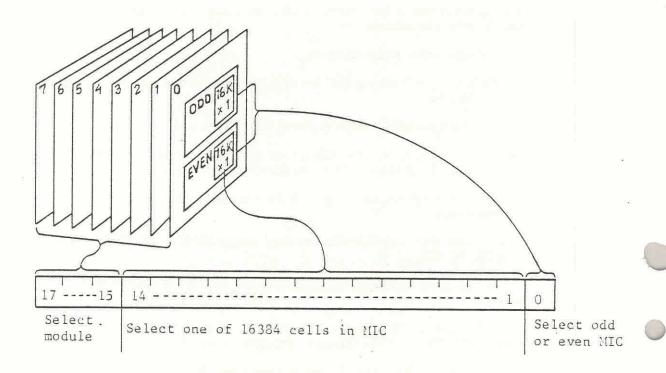


Figure 1.3: Use of Internal Bank Address

# 1.3.5 *Port (1142, 1143)*

A port is the entrance for a channel to a bank. A bank holds from 1 to 4 ports (A - D). A port consists of receivers for addresses and transmitters and receivers for data and control lines.

Up to 10 ports can be linked together (daisy chained) by a channel, i.e., a source can access up to 10 banks of storage.

A port consists of a data (1143) and an address (1142) module. A port carries out two major functions:

- Select a bank for the source and convert the channel address to a local bank address (1142).
- Generate a 5 bit control code to be attached to data during write and
  - check the data against the control code during read.

Each port defines the address range for a bank (1142). (Refer to Section 1.4.) Normally, the address range is the same as seen from the different ports, but for special purposes they will be different. One such example is the NORD-10/NORD-50 communication using BMPM as a common memory.

# 1.3.6 Channel

In the memory system a Channel is physically a pair of cables. One cable carries address and request signals while the other carries data and ready signals. With the word channel we will normally mean a 16 bit data channel. It is, however, possible to define a 32 bit channel consisting of one address cable and two data cables. Since all ports are 16 bits wide, two ports are required in that case. A 32 bits channel is illustrated in Figure 1.1.

All channels meet only one type of port. Hence, all sources must conform to the same *specifications* when requesting memory.

It is normally possible to connect several banks to one channel.

Every port has one input and one output connector for each cable, which allows daisy chaining of up to 10 ports. At the end of the chain a termination plug is mounted.

*Note!* The two banks in the same crate have their ports daisy chained internally. Figure 1.4 shows signals appearing on a channel.

### CHANNEL

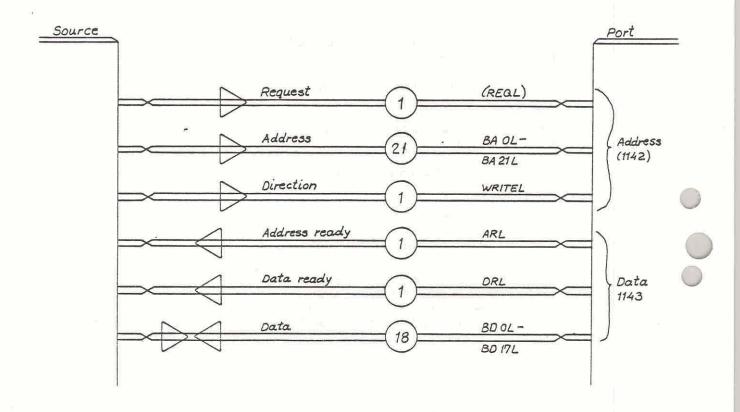


Figure 1.4: Channel Signals

Note 1: A 32 bits channel consists of one address cable and two data cables.

Note 2: Up to 10 ports may be chained to one source. The cables must be terminated in the last port.

Transmitters and receivers in the ports meet the RS-422 specifica-Note 3:

tions. (Refer to Appendix E.)

#### 1.3.7 Source

A Source can be any kind of electronic unit which is capable of communicating with BMPM over a channel. Examples of sources can be: NORD-10, NORD-10/S, NORD-50 or DMA interfaces. The sources must conform to the channel specifications (refer also to Figure 1.1).

# 1.3.8 Service Channel

The service channel consists of an error log module (1145) located in the BMPM, an error log I/O interface module (1146) located in the NORD-10/S input/output system and a cable connecting those two modules together. BMPM will thus be regarded as an I/O device as seen from NORD-10/S.

The error log module (1145) is logically divided into two parts, one serves the X-bank and the other the Y-bank. Each part contains an error log memory. All errors (single or multiple) occurring in the bank and detected by the ports (1143's), will be recorded in this special memory. The error log memory is organized as a 512 x 1 bit memory where the address is a pointer to the failing memory integrated circuit (MIC).

A thumb wheel switch located on this module defines the crate number (0 - 7).

The service channel can perform the following four functions:

1. (Action: 00), Refer to Figure 1.5.

Scan the error log memory for a failing bit. The error log memory address where the failing bit is found (pointer to the failing MIC) is transferred to a one-word scan register located in the error log I/O interface (1146).

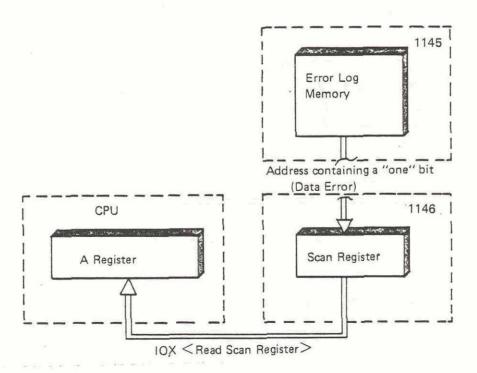


Figure 1.5: Scan Error Log

### 2. (Action: 01), Refer to Figure 1.6.

Scan the present ports for bank address range setting. This operation will also give the information whether the ports have detected an error or not.

Steps 1 and 2 are feed-back information from the BMPM and will be written into the scan register located in the error log I/O interface (1146). This will be transferred to the A register of the CPU by execution of an IOX <Read Scan Register> (IOX 750).

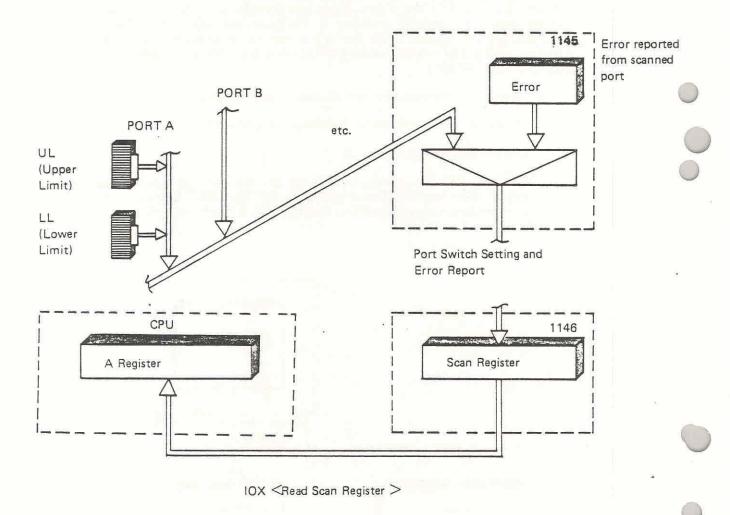


Figure 1.6: Scan Port Status

3. (Action: 10), Refer to Figure 1.7.

Set the content of the error correction control register (ECC) located in the data module (1143) of the ports.

(Action: 11), Refer to Figure 1.8.

Execute test access. A read or write operation can be simulated without a channel connected.

Steps 3 and 4 are control information for the BMPM. The control information is transferred from the A register in the CPU to the command register located in the error log I/O interface (1146) by executing an IOX <Load Command Register> (IOX 751). From this register the control information is sent to BMPM via the error log module (1145).

Information regarding detailed description of programming specifications and word formats is given in Chapter 3.

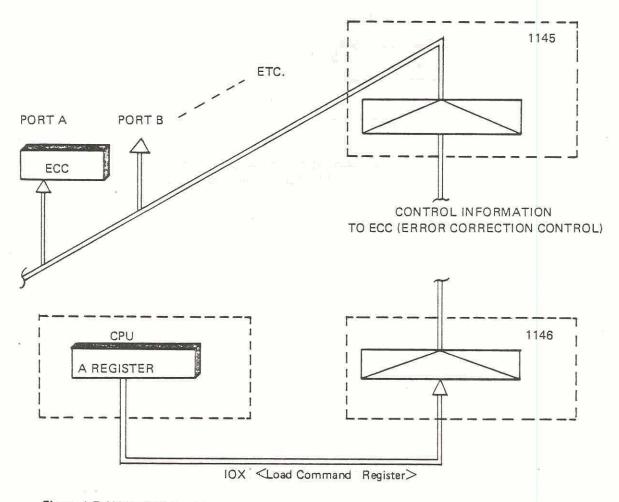


Figure 1.7: Write ECC Register

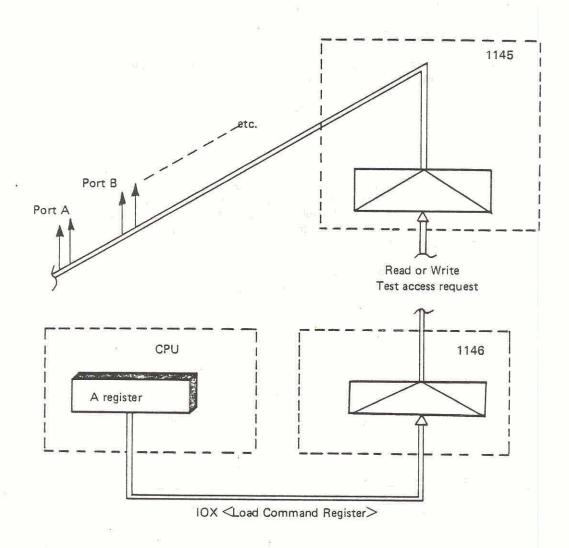


Figure 1.8: Execute Test Access

# 1.4 ADDRESSING

### 1.4.1 Address Conversion

The total address range of a channel is 2048K (21 address bits). Since each bank has a range of 256K, 8 banks are necessary to cover the complete range. Hence, the 3 most significant address bits could be decoded for bank selection.

However, since the minimum size of a bank is 32K (1 storage module) it is desirable that one has a resolution of 32K for selection. Hence, the 6 most significant bits are needed for decoding of 32K banks.

Straight decoding requires banks of equal size, so to permit banks of different sizes a more flexible method is required.

Each port has therefore a set of limit switches which define the address range the port will respond to.

One pair determines lower limit, LL, and one pair determines upper limit, UL. Each pair holds a two-digit octal number. Refer to the chart in Figure 1.9 for the corresponding address range.

When a request appears on the channel, all ports test the address against their limit switches. The request shall have access if:

LL ≤ Channel Address < UL

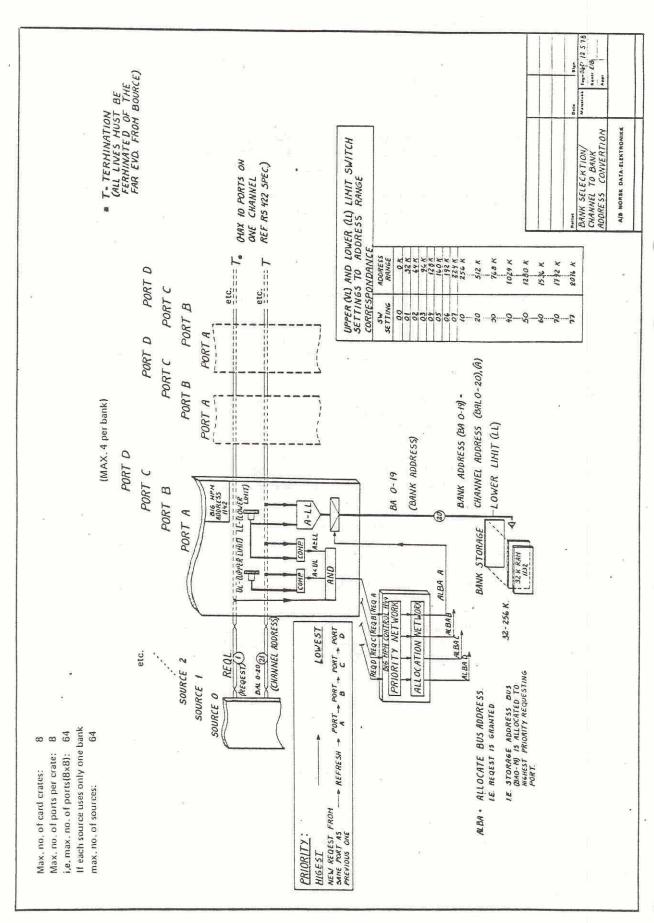
Inside all banks the address range is 0 - 256K, so the channel address must normally be transformed to an internal bank address which is offset by the lower limit (LL).

This transformation is performed at the port, and it is a simple subtraction:

Bank Address = Channel Address - Lower Limit

Accordingly, channel address equal to lower limit is directed into bank address 0.

Observe that two channels may access the same memory cell although their channel addresses are different. Address skew between channels should be avoided as much as possible since it makes the system very opaque both from a hardware and software point of view.



Fir. re 1.9: Bank Selection/Channel to Ba Address resion

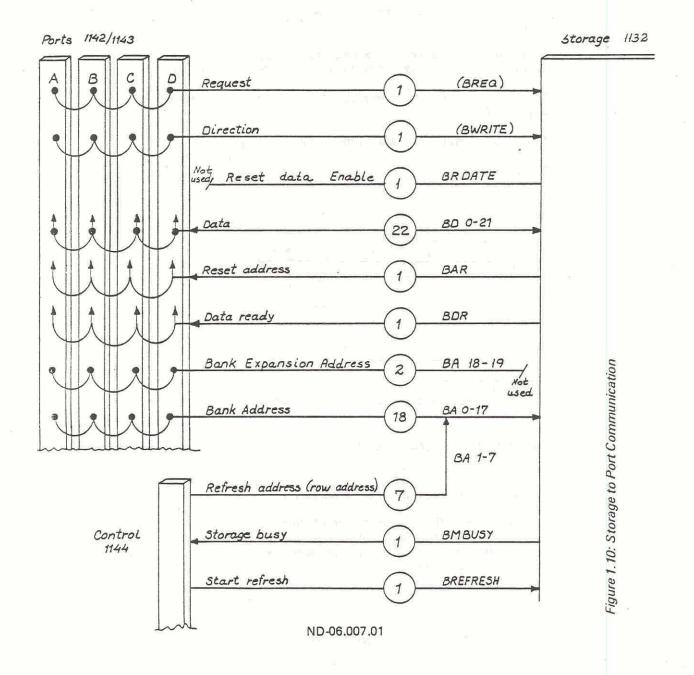
# 1.4.2 Bank Selection

For the following discussion refer to Figure 1.9.

When a source places an address on the bus together with a Request all the ports linked together by the channel will examine the channel address. The port (1142) that finds the channel address within its limits ( $LL \le A < UL$ ) will forward the request to the controller (1144).

Since the 4 ports associated with a given bank operate completely asynchronously with respect to each other, a priority network is required. The highest priority port will then be allocated to storage for one memory cycle. The bank address = (channel address — lower limit) will then be enabled onto the local bank address bus.

For further information regarding storage to port communication refer to Figure 1.10.



# 1.4.3 Interleave (Shifted Address)

To increase memory band width an interleave techique can be used. However, it is only meaningful in a pipeline system where a new request is issued before the previous is serviced.

In a less critical sense "interleave" is used to designate shifting of the address bits.

If the address is rotationally shifted one place to the right, bit 0 will receive position 20, bit 20 position 19 and so on. The effect is that all even addresses will seem to lie between 0 and 1024K while all odd addresses will appear as lying between 1024K and 2048K.

Two consecutive addresses (as seen from program) will therefore lie in different banks. The band width benefit expectation from this shifting is greatly exagerated when memory accesses are not pipelined.

However, interleaving is necessary when 16 bit and 32 bit channels are to communicate.

Two 16 bit words, at consecutive addresses, may be read as one 32 bit word when the two 16 bit words reside in different banks. (See Figures 1.12 and 1.13.)

The address shifting is performed by modifying the address cable as shown in Appendix C.

The concept may be illustrated in the following example.

#### Example:

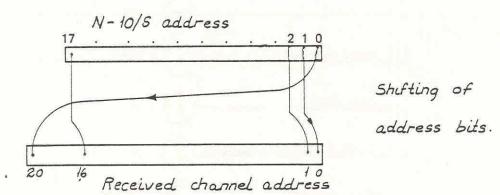


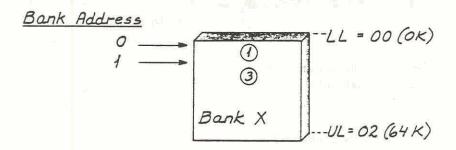
Figure 1.11: Shifting of Address Bits

Address bit number 20 is used to select a bank.

Assuming NORD-10/S is storing 4 words in memory in consequtive locations (as seen from NORD-10/S) the lowest bit on the issued address will toggle and alternating banks will be selected. If the issued addresses are 0, 1, 2 and 3, the received addresses would be:

Re	ceiv	/ec	Α	dress:	Word No.:		
0 0	0 0	0	0	0	0	1	
4 (	0 0	0	0	0	0	2	
0 0	0 0	0	0	0	1	3	
4 (	0 0	0	0	0	1	4	
6	etc.						

The following illustration will show how the four words will be stored.



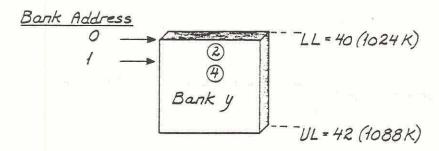


Figure 1.12: The Banks as Seen from NORD-10/S

When NORD-50 reads those 4 NORD-10/S words, it regards the same two banks as one bank with double word length. The 4 16 bit words will then be read as two 32 bit words.

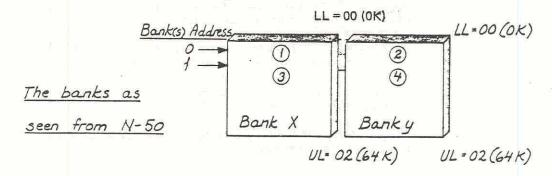


Figure 1.13: The Banks as Seen from NORD-50

# 1.5 DATA

### 1.5.1 Format

As seen from the source (NORD-10/S), the standard BMPM data width is the 16 bits associated with 2 odd parity bits, one for lower, one for upper byte.

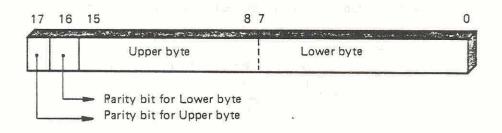


Figure 1.14: Channel Data Format

This makes the BMPM compatible with the former model.

# 1.5.2 Data Protection

Data is protected by means of a 5 bit control code. Error correction is therefore standard in BMPM.

### 1.5.2.1 General about Error Detection and Correction

In an error correcting memory 5 parity bits are stored in addition to the normal 16 data bits. These 5 bits are parities formed by the data bits in a unique way. Each port has its private error correction network in order to load the common memory resources as little as possible. The two source-generated parity bits are skipped and 5 new control bits are generated. The port could have checked the parity of write data but hardly informed the system in a proper way. It could have reported "Hey, I have seen an error", but it is impossible for the system to decide what action to take on such information. The reason for error would probably be a faulty cable, which would be detected by reading anyhow.

Although many people find error correction obscure, the practical treatment is quite straight forward and is shown in Figure 1.21.

For the sake of clarity, generation and checking are shown separately although they actually have common circuitry.

The following 5 parities serve as control bits and are generated at each write access.

```
\begin{array}{c} \text{C0} = 0 \oplus 1 \oplus 2 \oplus 3 \oplus 4 \oplus 6 \oplus 8 \oplus 10 \\ \text{C1} = 0 \oplus 2 \oplus 4 \oplus 7 \oplus 9 \oplus 10 \oplus 12 \oplus 14 \\ \text{C2} = 1 \oplus 2 \oplus 5 \oplus 7 \oplus 11 \oplus 12 \oplus 15 \\ \text{C3} = 3 \oplus 4 \oplus 5 \oplus 6 \oplus 7 \oplus 13 \oplus 14 \oplus 15 \\ \text{C4} = 8 \oplus 9 \oplus 10 \oplus 11 \oplus 12 \oplus 13 \oplus 14 \oplus 15 \end{array}
```

They are stored together with the 16 data bits in memory.

When data is read the parity of the 21 bit word is checked. If no errors have been introduced all parities are unchanged which gives the error code 00000 = GOOD. The 5 bits in the error code are called syndrome bits. If GOOD does not show up there must be a single bit or a multiple bit error.

How can these five bits really point out the faulty bit? The reason is that two different data bits never contribute in the same manner in *all* parities. For example, bit 2 is a part of C0, C1 and C2. If bit 2 has been changed (inverted) since generation C0, C1 and C2 will also be inverted. Hence, the error code 00111 is produced instead of the GOOD code 00000 which means no bit changed. Reference to Figure 1.15 will show that, in fact, 00111 is the error code for bit 2. Since error in bit 1 changes C0 and C2, 00101 is the error code for bit 1, etc.

It is important to note that only error in one bit at a time is assumed.

Since 21 bits can produce only 21 single error codes the remaining codes must be caused by *errors in some combination of two or more bits.* Those errors cannot be corrected and are therefore fatal to the system.

A decoder is used to decide which of the 31 possible errors has occurred (see Figure 1.15).

In case it is a single bit error the correspondig bit is inverted by an EXCLUSIVE-OR gate.

In case a multiple error is detected a FATAL ERROR signal is generated. This signal inverts the otherwise correct parity bit 16.

Accordingly, multiple errors force a parity error which is detected by a normal parity check in the source. Single bit errors are (of course) not reported to the source. However, all errors are written in the error log (1145).

The theory for error correction is also covered in Section II.6 of the manual "NORD-10/S Functional Description (ND-06.009).

	S4	S3	S2	<b>S</b> 1	S0	No Error	Single Data Error	Single Error Error	Multiple Errors
0 1 2	0 0 0	0 0 0	0 0 0	0 0 1	0 1 0	Good		C0 C1	
3 4 5	0 0 0	0 0 0	0 1 1	1 0 0	1 0 1		E0 E1	C2	
6 7 10	0	0 0 1	1 1 0	1 1 0	0 1 0		E2	C3	MEO
11 12 -13	0 0	1 1 1	0 0	0 1 1	1 0 1		E3 E4		ME1
14 15 16	0 0 0	1 1 1	1 1 1	0 0 1	0 1 0		E5 E6 E7		
17 20 21	0 1 1	1 0 0	1 0 0	1 0 0	1 0 1		E8	C4	ME2
22 23 24	1 1 1	0 0 0	0 0 1	1 1 0	0 1 0		E9 E10 E11		
25 26 27	1 1 1	0 0	1 1 1	0 1 1	1 0 1		E12		ME3 ME4
30 31 32	1 1 1	1 1 1	0 0	0 0 1	0 1 0		E13		ME5
33 34 35	1 1 1	1 1 1	0 1 1	1 0 0	1 0 1		E15		ME6 ME7
36 37	1	1	1	1	0				ME8 ME9

Figure 1.15: Syndrome Decoding

### 1.5.2.2. Write

During a store operation the port (1143) will receive the data in the format as indicated in Figure 1.14.

The two parity bits are not processed in any manner by the port. Based on the 16 data bits a 5 bits control code is generated and sent to memory with the data. This is illustrated in Figure 1.16.

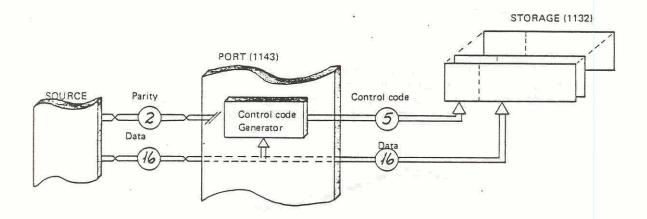


Figure 1.16: Data Flow on Write

Note! Addressing and bank selection is not covered here.

### 1.5.2.3 Read

During a read operation the addressed data is received from the selected storage by the port (1143). Working on the 16 data bits, a new control code is generated and compared with the one stored with the data. If they are equal, the data is accepted as Good data.

The difference (exclusive OR) between the "new" and the "old" control code is called the syndrome. If the syndrome is equal to 0, the data is Good. As the control codes are checked, two parity bits are generated in the port (1143) and sent to the source with the data. See Figure 1.17.

# 1.5.2.4 Single Data Error

If a single data bit is failing, the 5 bits syndrome (exlusive OR of "new" and "old" control code) will be different from zero ( $\neq$  0) where the syndrom (code) is a pointer to the failing bit. The failing bit is corrected (inverted) and the correct parity bits will be generated and sent to the source with the correct data. The syndrome ( $\neq$  0) will be sent to the error log module (1145) to indicate where and what type of error has occurred. See Figure 1.18.

# 1.5.2.5 Single Control Code Error

Using the same type of memory for storing the control code, a single bit can also be failing in the code. The syndrome (code) will also here be a pointer to which control bit that was failing. No data need be corrected for this type of error. The synfrome is sent to the error log module (1145).

# 1.5.2.6 Multiple Error

A multiple error is a type of error where more than one bit is failing within the 21 bits field. In this case the data cannot be corrected. To indicate that the data is not correct, the parity bit for the lower byte is forced false (BDL 16). The syndrome is here also sent to the error log module (1145). See also Figure 1.19.

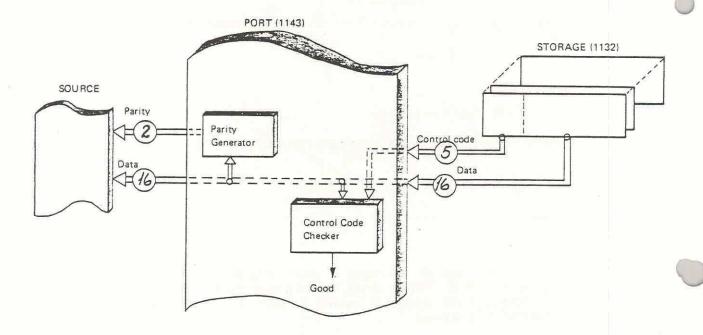


Figure 1.17: Reading Data (No Error)

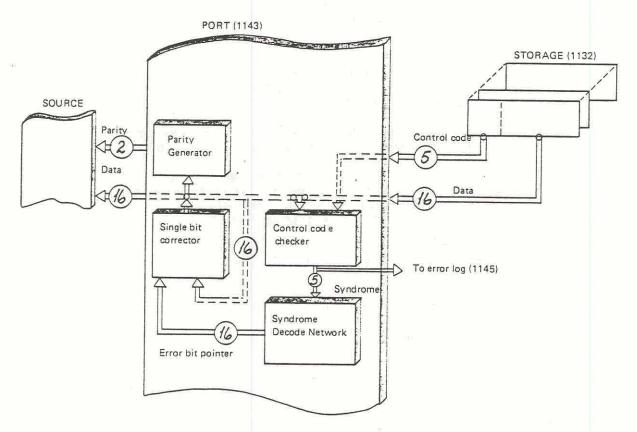
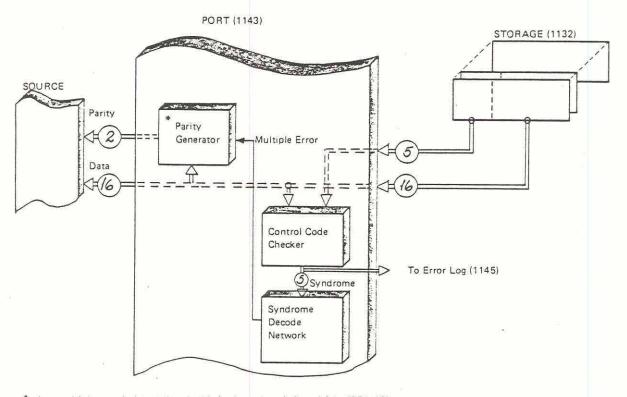


Figure 1.18: Reading Data (Single Error)



<sup>\*</sup> when multiple error is detected parity bit for lower byte is forced false (BDL 16)

Figure 1.19: Reading Data (Multiple Error)

# 1.5.3 Control and Error Reporting

We have already discussed that single bit errors are detected and corrected internally in the BMPM. If multiple errors occur, the data cannot be corrected, and parity bit for lower byte is forced false.

If NORD-10/S is the source, an internal interrupt is generated which forces the CPU to level 14. It is, however, of interest to be aware of a single bit error also.

Over the service channel and the I/O system a failure in BMPM may generate interrupt to level 13 at the time of occurrance. By selecting the proper bit mask for the error correction control register (ECC), we can decide whether interrupt is disabled, single bit errors or multiple errors should generate interrupt.

At the time of the failure (interrupt) the syndrom is sent to the error og (1145). By reading the error log memory the system will know what type and where the error occurred.

See detailed description in Chapter 3.

Refer also to Figure 1.20 for further details.

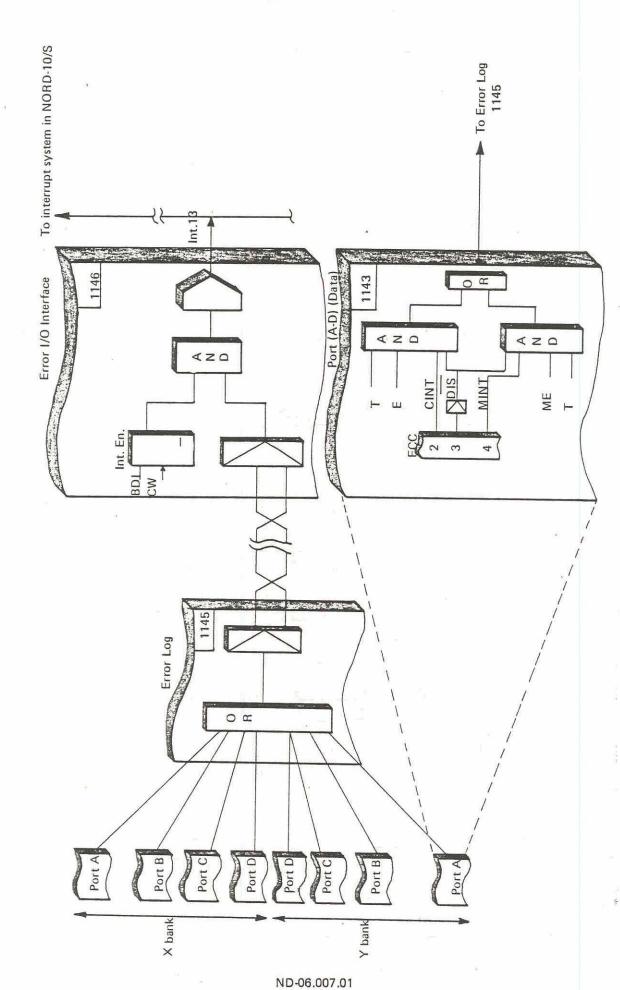


Figure 1.20: I/ O Interrupt System in NORD-10/S

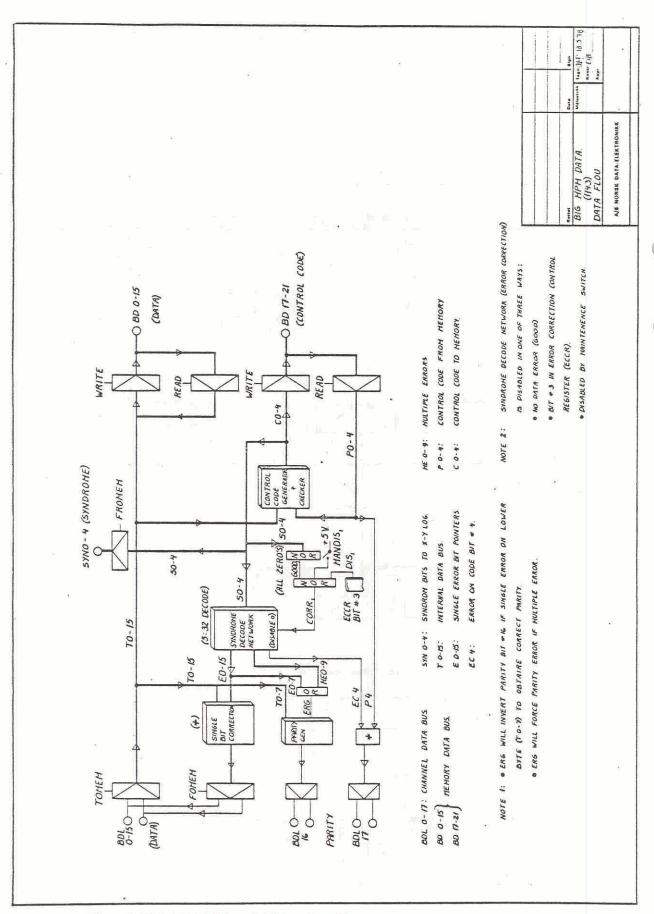


Figure 1.21: Big MPM Data (1143) — Data Flow

# 2 MEMORY CHANNEL SPECIFICATIONS

A 16 bits channel consists of one address cable, one data cable and two termination plugs. A 32 bit channel consists of one address cable, two data cables and three termination plugs. The specifications herein concern a 16 bit channel, unless otherwise noted. The channel source must fulfill the electrical and logical requirements stated below.

# 2.1 HARDWARE SPECIFICATIONS

# 2.1.1 Signal Standard

The channel signals are carried on differential lines conforming to CCITT-V.11 or RS-422 standard. See Figure 2.1 and Appendix E for further details.

Recommended line drivers and receivers are: AMD 26LS31 and AMD 26LS32 or equivalent.

### 2.1.2 Cable

Cables may be 120 ohm twisted pairs or flat cable. Maximum total length of a channel is 15 m. This limitation is due to requirement of signal quality at 10M baud.

# 2.1.3 Connector

Cables may be connected either direct to backwiring posts or via an optional plug panel. The backwiring connectors may be one of the following:

For twisted pairs;

BERG 65051-011, 2 x 26 position latch housing

BERG 47712, discrete female sockets

For flat cables;

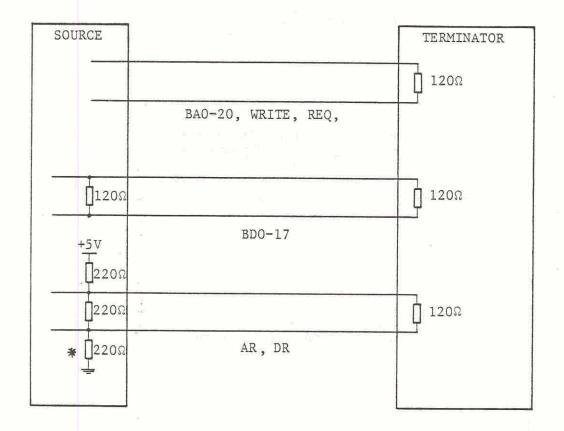
3M 3307, 2 x 25 position socket connector

3M 3404, keying header

### 2.1.4 Termination

All cables longer than 1 meter or between two cabinets shall be terminated at the end of the daisy chain (far end from source). Each differential pair shall be linked with a 120  $\operatorname{\mathfrak{K}}$  resistor mounted on a suitable connector (see Appendix C.6).

On the source module of the data cable further termination has to be performed. The termination scheme is summarized in Figure 2.1.



<sup>\*</sup> Provisional. May be subject to change.

Figure 2.1: Termination of Memory Channel Signals

### 2.2 SIGNAL DESCRIPTION

Complete signal lists are found in Appendix C.2 and C.3. Below is a description of signals as seen from the source.

For the following discussion refer also to Figure 1.4.

# 2.2.1 Address Cable Signals

BA 0-17:

18 output address signals for normal NORD-10 address

range

BA 18-20:

3 output address signals for extended address range

REQ:

output pulse to request access

WRITE:

output signal indicating data direction. WRITE = true

means write to memory. WRITE = false means read from

memory.

# 2.2.2 Data Cable Signals

BD 0-15:

16 bi-directional data signals

BD 16-17:

2 bi-directional signals indicating odd parity of lower and

upper byte respectively

AR:

Input signal, address ready. Indicates that the request is

accepted by a bank and that another request may be

issued.

DR:

Input signal, data ready. Indicates that write data is

accepted by bank or that data read from memory is valid

on data lines.

### 2.3 CHANNEL TIMING

Timing diagrams and requirements are shown in Figures 2.2, 2.3A and 2.3B.

# 2.3.1 Access and Cycle Times

Access time at the port, measured as time between REQ and DR, assuming no latency.

Write:

200 ns maximum

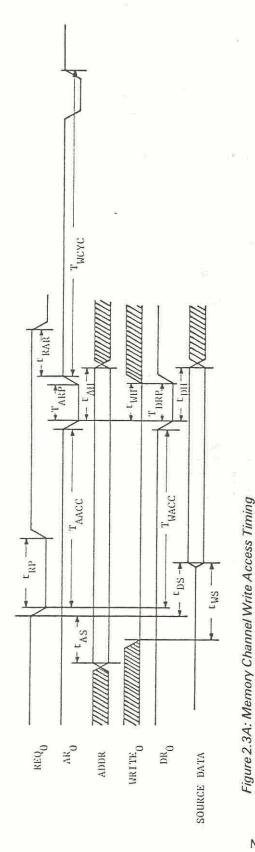
Read:

450 ns maximum if data is good 500 ns maximum if data is corrected

Description:	Name:	Min.:	Туре	:Max.:	Unit:
Address set-up time	<sup>t</sup> AS	0			
Write data set-up time	t <sub>DS</sub>	-75			ns
Write command before write data enabled	tws	70			ns 1)
Write data hold time	tDH	0			
Write command hold time	<sup>t</sup> WH	0			
Address hold time	<sup>t</sup> AH	0			
Address ready pulse width	TARP	60	90		ns
Request pulse width	<sup>t</sup> RP	50		∞	ns
Data ready pulse width	TDRP	70	100		ns
End of address ready to next request	<sup>t</sup> RAR	0			
Access time from Request to Address Ready	TAACC			200	ns 2)
Write data access time	T				
Maine and a single	WACC			200	ns 2)
Write cycle time		380			ns
Read command set-up time	<sup>t</sup> RS	-30			ns
Read data disabled after Read Command false	TRDIS	25		70	ns 1)
Read data valid before Data Ready	TOVAL	0			
Read data access time (no error)	TRACC			450	ns 2)
Read data access time (1 bit corrected)	TRACC			500	ns 2)
Read cycle cycle time	TRCYC	380			ns

Figure 2.2: Memory Channel Timing Specifications Measured at Port Terminals

- Write signal should be put true as soon as possible after each read access in order to disable data driver at port.
- Access times in case there is no waiting time due to previous accesses in the bank. Minimum cycle-time will limit write access time.



T RD1S -DVAI -- KAK A. ARIT t RP LRS-WRITE<sub>0</sub> ADDR PORT DATA

Figure 2.3B: Memory Channel Read Access Timing

Cycle time at the port, measured as time between two AR. Next REQ at trailing edge of AR.

Write:

380 ns maximum in same address. May decrease to 300 when acces-

sing alternate modules.

Read:

380 ns maximum.

## 2.3.2 Latency

Due to different activities in the bank there may be a certain waiting time in addition to the regular access time. Maximum waiting will occur when two succeeding accesses attend the same memory integrated circuit.

- 1. Maximum waiting time when one higher priority port has access: 480 ns.
- 2. Maximum waiting time when two higher priority ports have access: ∞.
- 3. Maximum waiting time due to refresh: 520 ns.
- 4. Maximum waiting time due to previous access from same port: 180 ns.

The above numbers arise from worse case conditions which are seldom experienced.

### 2.3.3 Timeout

Missing answers AR and DR from a channel may be due to "memory out of range" or "memory inhibit" caused by power fail in a bank. It is therefore mandatory that the channel source has a timeout mechanism.

Minimum length is 10 vs.

### 2.4 32 BITS CHANNEL

A 32 bits channel is obtained by accessing two banks simultaneously from one source. The addresses and requests are the same in both banks as the address cable is common. Two data cables will each supply half of the 32 bits word (see Figure 2.4). As the source receives a double set of ready signals, which may be out of phase, there must be a provision to wait for the slowest response.

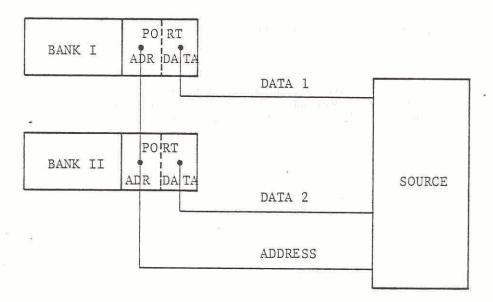


Figure 2.4: Two 16 Bit Banks Accessed as One 32 Bit Bank

### 2.5 INDICATORS

The data modules have two red indicators which normally shall be off. The upper warns that an error is detected. The lower warns that the error correction network is disabled either manually or by program.

The address modules have three yellow indicators of general information value. The upper indicates that the last request was accepted by the port. The middle indicates that the request was granted by the bank control. The lower indicates that a storage module has answered. These indicators light up until cleared by the next request on the channel.

### 2.6 ADDRESS RANGE SWITCH

The address range which a port shall respond to is determined by two sets of switches on the address module.

The upper set determines upper limit in 32K increments. The lower set determines lower limit in 32K increments. See Section 5.1 for further details.

## 2.7 DISABLE ERROR CORRECTION SWITCH

The data modules have a switch for manual disable of error correction. Operating the switch will also clear the error indicator. The switch is intended for maintenance purposes and failure will result if operated during an access.

# 3 SERVICE CHANNEL SPECIFICATIONS

### 3.1 GENERAL

# 3.1.1 *Error Log (1145)*

Each port contains an error checking and correction circuitry which detects all single bit and alot of multiple bit errors. (Turn to the manual "NORD-10/S Functional Description" for further information.)

When an error is detected by a port, the 5 *syndrome* bits which identify the error are forwarded to the *error log* module. This module has a 512 x 1 bit log memory. The 9 bits address to the log is composed of the following parts:

8	7	6	5	4	3	2	1	0
A17		A15	A0	S4				S0

S 0-4:

Syndrome bits indicate failing data bit (or eventually

multiple error)

A0:

Address bit that indicates even or odd block on failing

module

A 15-17:

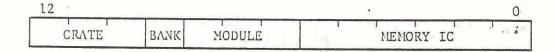
Address bits that identify the failing module

It is apparent that this 9 bit address is sufficient to point out the failing memory integrated circuit (MIC). In a 256K memory bank there are 336 MIC's hence every MIC has an associated cell in the log. Every time an MIC fails the corresponding cell is set to 1. The log will only indicate fail or not fail and not the number of fails.

Each time the log is read it is cleared and it is possible to obtain a good error statistic by reading the log many times.

# 3.1.2 Error Log I/O Interface

The error logs are not ordinary working memory and therefore they need a dedicated interface to read them. The interface is a normal I/O module obeying a number of IOX instructions. An IOX command initiates scanning of the error log in a bank, bit for bit. If 1 is hit on the address together with bank and crate number it is returned to the CPU. This information uniquely identifies all MIC's in a 16 bank system.



The log channel also serves other service purposes which described later. Refer also to Figure 3.1.

## 3.2 PROGRAMMING SPECIFICATIONS

## 3.2.1 Device Number

Fixed wired device numbers are 750 - 753.

3.2.2 
$$Interrupt Level = 13_{10}$$

Interrupt is only caused by error conditions, i.e., status bit 1 and status bit 4 = 1.

### 3.2.3 | Ident Code = 5

Fixed wire.

# 3.2.4 Load Command Register = IOX 751

The command is loaded but the result depends on status bit 2.

If status bit 2 = 0 (scan not active), clear status bit 3 (ready for transfer). Execute command.

If status bit 2 = 1 (scan active), abort scan. Such termination is possible, but should be done only when absolutely necessary.

The format is:

15	14	13	12		9	8				0
ALL	ACT	ION		BANK			1	I	ATA	

Bits 0-8:

Data associated with actions 10 and 11. Unused in action

00 and 01 (see Section 3.3).

Bits 9-12:

Bank address where bits 10-12 are crate numbers (see

Section 3.4.5). bit 9 = 0 means X-bank and bit 9 = 1

means Y-bank.

Bits 13-14:

Action caused in selected bank.

14 13

0 0 Initiate scan of error log

0 1 Initiate scan of port status

1 0 Load error correction control register

1 1 Execute a test access

The actions are further described in Section 3.3.

Bit 15:

Address all banks.

If bit 15 and bit 14 = 1 perform action in ALL banks regardless of bits 9-12.

Otherwise use bits 9-12 as bank address. (The purpose is to allow setting of all error correction control registers in one instruction.)

## 3.2.5 Read Scan Register = 10X 750

Reading of this register is only meaningful after initiation of a scan (action 00 or 01 of command register).

If status bit 2 = 1 (scan active): Clear status bit 3 (ready for transfer). Read register.

If status bit 2 = 0 (scan not active): Read register with data bits 0-8 = 0.

The format is:

15	14	13	12		9	8			0	
CRATE	SCAN	ERR			T	T			1	
PRES	ACTIVE	STAT		BANK			DA?	ГА		

Bits 0-8:

Data resulting from actions 00 and 01 (see Section 3.3).

Bits 9-12:

Bank address where bits 10-12 are crate numbers bit 9 = 0 means X-bank and bit 9 = 1 means Y-bank. (These bits are the same as loaded by the last command.)

Bit 13:

Error status.

During a scan of port status (action 01) the bit indicates whether there has been an error at the port. Used for further identification of error interrupts (see Section 3.3). Bit 13 = 1 during scan of error log (action 00).

Bit 14:

Scan active.

If bit = 1 the addressed bank is busy scanning its error log or port status. The bit is identical to status bit 2.

Set: Initiate scan by action 00 or 01 of command register.

Clear: At end of scan. Change of command register (incorrect!).

Bit 15:

Crate present.

Use to separate missing crates from present crates with empty error log.

Set: The crate addressed by last scan command is present.

Clear: The crate addressed by last scan command did not answer.

# 3.2.6 Read Status Register = IOX 752

Since the interface is always ready for transfer in less than 100  $\nu$ s neither of the data instructions (IOX 750-751) can cause interrupt. The only sources of interrupt are memory errors detected by the ECC system at each port.

The meaning of the status bits are:

Bit 1:

Error interrupt enabled.

Set: Control word bit 1 = 1.

Clear: Control word bit 1 = 0. Control word bit 4 = 1.

Serviced IDENT PL 13. Master Clear.

Bit 2:

Scan active.

If bit = 1 the addressed bank is busy scanning. The bit will be 1 during the whole scan period even when valid data is transferred to the interface. Hence it is *not* inverse

of status bit. 3.

Set: Initiate scan by action 00 or 01 of command

register.

Clear: At end of scan. Change of command register.

Bit 3:

Ready for transfer, RFT. RFT does not cause interrupt.

Set: Status bit 2 = 0. (Always ready when scan is passive. Status bit 2 = 1 and valid data present in interface. Control word bit 4 = 1 (device clear). Master Clear.

Clear: Status bit 2 = 1 and valid data not found yet (but

investigation is in progress).

Bit 4:

Memory error.

An OR function of the error signals from all ports of all banks. The individual contributions are enabled by error correction control register.

Set: Single or multiple memory error detected and enabled for interrupt at the port. Control word bit 3 = 1 (set to test mode).

Clear: Scanning of port status of the failing bank. Test mode is cleared by:

Device Clear

Services IDENT PL13

Master Clear.

## 3.2.7 Load Control Register = IOX 753

Bit 1:

Enable interrupt on memory error

Bit 4:

Device clear.

Set RFT = 1, Terminate scan, Clear Test mode.

Device clear has no influence on memory error signals which must be cleared by scanning of port status.

### 3.3 DETAILED DESCRIPTION OF COMMANDS

Besides handling error interrupt, the interface provides execution of four command actions.

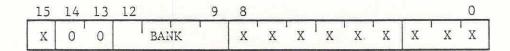
The actions are either *direct* or *scan*. When a scan is initiated by a command it must be allowed to terminate before another command is started.

To find out when the scan register content is valid, status bit 3 must be read repeatedly. (Make a loop with error exit after 100  $\nu$ s without finding status bit 3 = 1.)

#### 3.3.1 Action 00: Initiate Scan of Error Log

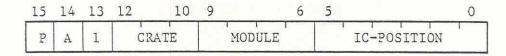
See also Figures 3.3 and 3.4.

Load the command register with the following content:



The interface will be set to scan mode (status bit 2 = 1. The addressed log module now searches through its error log and looks for error bits set. When status bit 3 changes to 1 the location of the error is transferred to the scan register. Eventually end of scan is reached without any errors detected.

The format of the error information is:



The information is here divided in a slighly different manner compared to what is stated in Section 3.1.1.

IC POSITION:

6 bits code showing the failing memory IC. Conversion between code and physical position is stated in Appendix

MODULE:

4 bits code showing the slot of the failing memory module. Conversion between code and physical slot is stated in Appendix B.2..

CRATE:

3 bits code identifying the crate number displayed on log module in slog 18 or 19.

Scan active.

A = 1: More data will come

A = 0: End of scan. A new command is required to start another scan. Value of MODULE and IC POSITION is

insignificant. (Bits 0-8=0).

P

Crate present.

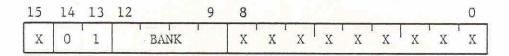
P = 0: Addressed crate does not answer. Scan aborted.

P = 1: Crate hale and hearty.

# 3.3.2 Action 01: Intiate Scan of Port Status

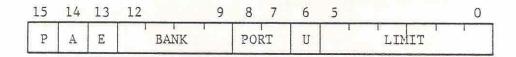
See also Figures 3.6 and 3.7.

Load the comand register with the following content:



The above will put the interface in scan mode. The addressed bank will look over its ports and report which are there and what their address ranges are.

The format of the port status is:



LIMIT:

Value of the range limit on thumb wheel switches of port

address module.

U:

Upper limit if U = 1.

Lower limit if U = 0.

PORT:

2 bits code indicting port. 0 - 3 means A-D respectively.

Appendix B.3 shows in which slots the respective range

switches are found.

BANK:

4 bits code constituted by one bit (9) indicating X or Y

and 3 bits (10-12) identifying crate number displayed on

log module in slot 18 or 19.

E:

Bit telling whether an error signal has been issued from

the current port.

A, P:

The meaning of these bits are the same as stated in

Section 3.3.1: Action 00.

Timing diagram for SCAN operation (ERROR LOG and PORT STATUS) is given in Figure 3.9 and 3.10.

# 3.3.3 Action 10: Load Error Correction Control Register, ECC

Each port has an error correction control register. It is similar (not equal!) to the corresponding CPU register for local memory access.

The format of the command is:

15	14	13	12		9	3					5	4	3	2	1	0
ALL	1	0		BANK		D	T	CPO	T DRT	B T	A	MUL EN	DIS	COR EN	TST 15	TST

TST0:

Invert polarity of correction code bits 0 and 1, which

correspond to error in data bit 0 (see below).

TST15:

Invert polarity of correction code bits 2, 3 and 4, which

correspond to error in data bit 15 (see below).

COREN:

Enable error interrupt on single bit errors.

DIS:

Disable error correction and clear error indicator on data

card (same function as the disable switch).

MULEN:

Enable error interrupt on multiple errors.

PORT ABCD:

4 bits port address where each port has a corresponding

bit. This allows access to all ports simultaneously.

BANK:

Bank address as described in Section 3.3.2.

ALL:

Address all banks regardless of the bank bits 9 - 12 if ALL = 1. ALL affects only the bank address not the port

addresses (which may be accessed simultaneously).

Evamole:

To enable interrupt on multiple errors in the complete system simultaneously, the following command word is

suitable:

140760

Note! At power-up and after a power-fail the ECC

registers are set to 0.

### 3.3.3.1 Test Error Correction

Error in data bit 0 is reported by means of the code 00011<sub>2</sub>. A forced inversion of bits 0 and 1 in the correction bits of otherwise correct data will exhibit error in data bit 0. At the same time a forced inversion of correction bits 2, 3 and 4 produce error code 11100 which accuses data bit 15 of being faulty.

But, what will happen if both the correction bits are inverted when written into and read from memory? Since two inversions are equivalent to no inversion, data will appear faultless. Thus, only the writing or the reading port should be in test mode in order to simulate errors.

### 3.3.4 Action 11: Execute Test Accesses

It is possible upon command to produce test requests at a port without attaching a channel.

The format of the command word is:

16	15	14	13	9	8			5	4				0
ALL	1	1	BAN	K	D	POI	B RT	Α	W	X	Х	Х	Х

PORT, BANK and ALL work in the same manner as described in the previous Section, 3.3.3.

If W = 1 a write access results and if W = 0 a read access results.

## 3.3.4.1 Address and Data of Test Accesses

How much information that can be drawn from a test access depends on whether the cable is connected or not.

When neither an address nor a data cable are connected the port will experience address = 777777 and write data = 0. The range switches must therefore be:

Lower limit: ≥7 Upper limit: ≤ 10

In order to forward the request to the bank. By means of the test mode (independantly set for each channel) it is possible to put an "error label" on selected data words. The labelled data word can be traced from and to specific ports. Furthermore, it is possible to follow the produced error to the log and to the I/O and interrupt system.

Even if the channel is connected to a physical source which sends out an unpredictable data and address, some information may still be extracted.

First, initialize the bank with the correct correction codes via an appropriate port. Then set test bit 0 or 15 on the port under test and perform the test access. It is immaterial what address and data is used. If this access produces error interrupt and correct the error code in the log very much of the port hardware works properly.

### 3.4 HARDWARE SPECIFICATIONS

The service channel is physically one cable linking all Error log modules to an I/O interface module. It has nothing in common with the ordinary memory channels. The main task is to read the log memories, but it serves also other purposes as described in Section 3.3.

### 3.4.1 Electrical Standard

The channel carries 20 uni- and bi-directional differential lines conforming to RS-422 standard.

Cables may be 120  $\Omega$  twisted pair of flat cables. For further details refer to Section 2.1.

### 3.4.1.1 Termination

All cables longer than 1 meter (or between cabinets) shall be terminated at the end of the daisy chain by a plug with 120 ohm resistor between each differential pair.

Bi-directional and input signals are also terminated on the interface module.

# 3.4.2 Service Channel Signals

A complete plug list is found in Appendix C.4. For the following discussion refer to Figures 3.1 and 3.2. The meaning of the signals are:

B0-8:

9 bi-directional data signals.

B9-14:

6 output command signals.

B15:

1 bi-directional control/status signal

LIOX:

Output signal indicating that command and data are valid from interface at output command. Handshake signal for

LDRY during input of scan data.

ACT:

Input signal. Answer from addressed module that it is

present and is active with a scan.

LDRY:

Input signal. "Data Ready" signal sent from active module telling about valid scan data. It shakes hands with LIOX in

scan mode.

LINT:

Input signal warning that an error is detected "some-

where". Causes interrupt if enabled.

# 3.4.3 Channel Timing

Timing diagrams and requirements are shown in Figure 3.1 to 3.12.

### 3.4.4 Error Indicator

Each LOG module has a red indicatgor lamp internally connected to the LINT signal. Indicates that an error has been detected in the crate. It is cleared by scanning of port status.

### 3.4.5 Crate Number Switch

The LOG module has a ten position thumb wheel switch defining a 3 bit crate number, 0-7. Position 8-9 puts the module off-line. The number is used for maintenance identification and is not required for normal memory accesses.

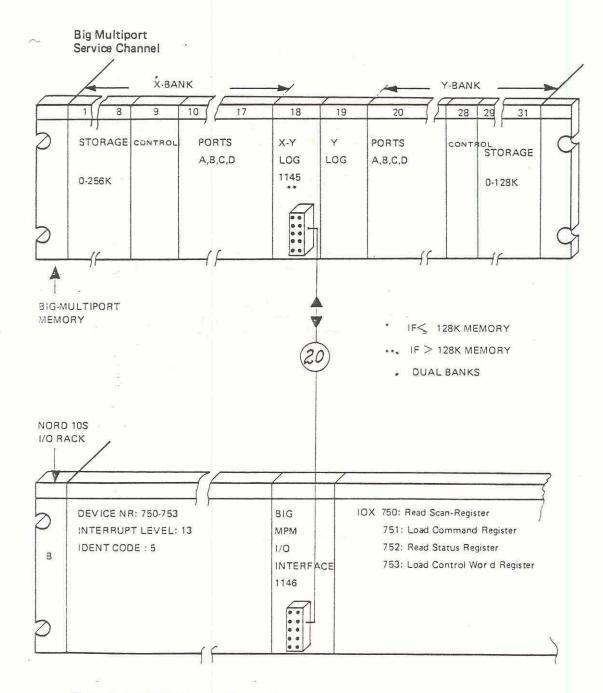


Figure 3.1: BMPM Service Channel

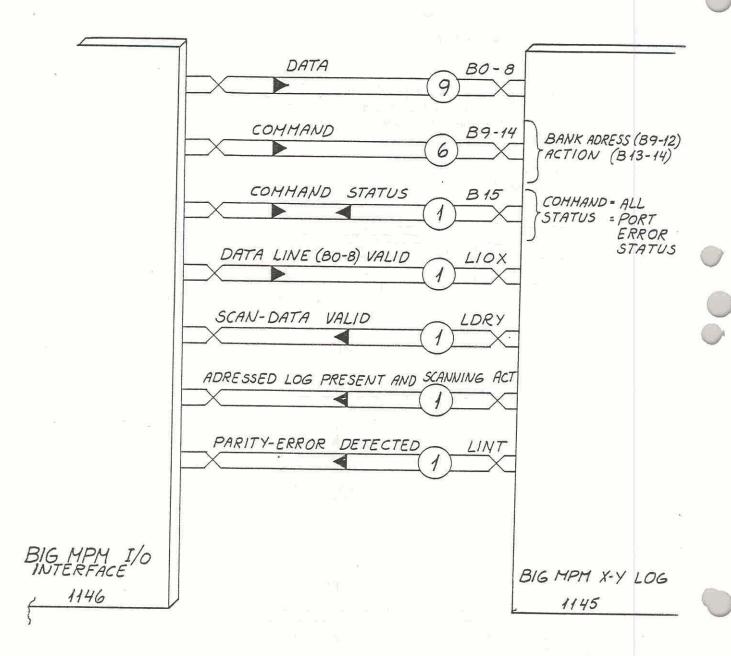
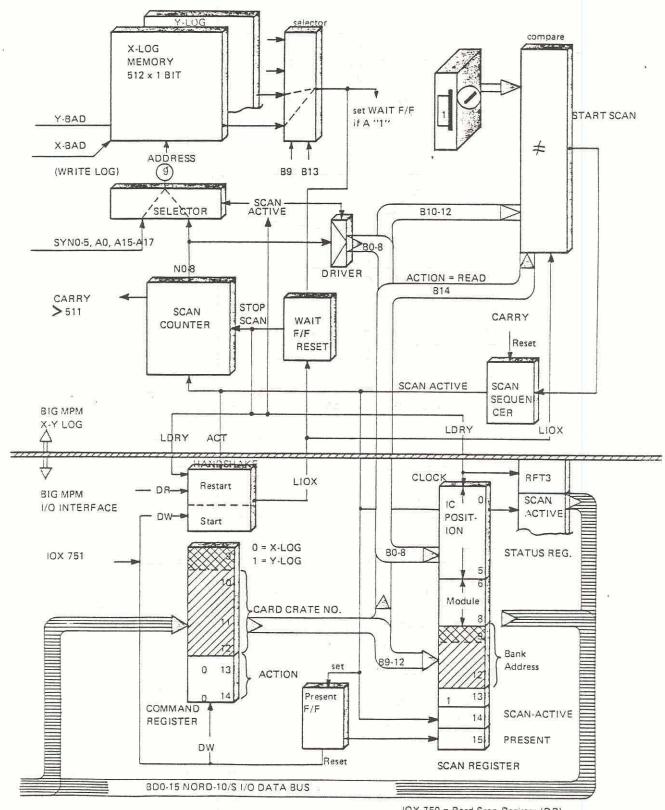


Figure 3.2: I/O Module to Error Log Module Communication



IOX 750 = Read Scan Register (DR)

IOX 751 = Load Command Register (DW)

IOX 752 = Read Status Register

Figure 3.3: Scan of Error Log — Block Diagram

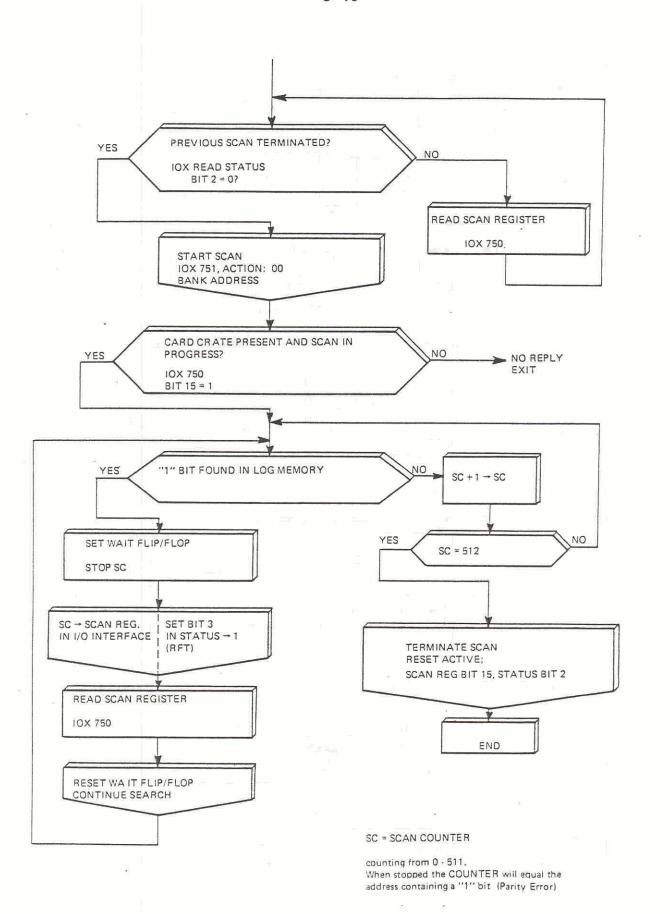


Figure 3.4: Action: 00 Scan of Error Log — Flow Diagram

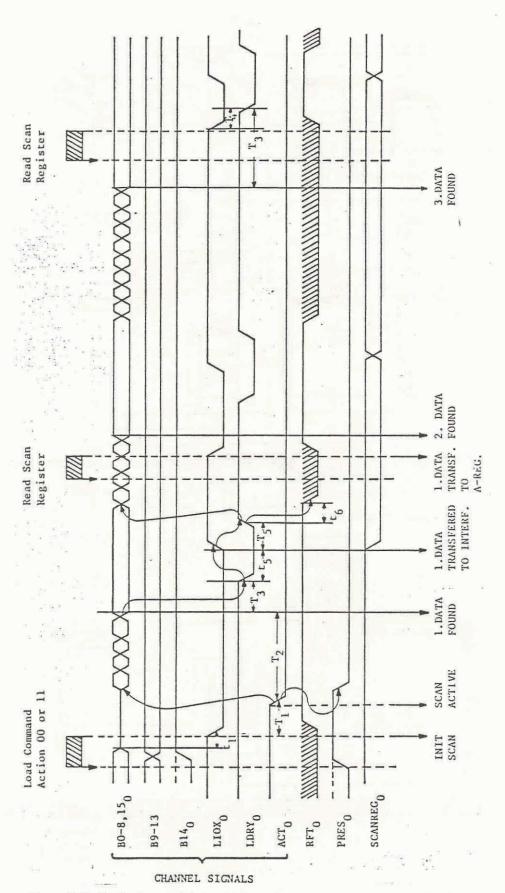


Figure 3.5: Beginning and Continuation of a Scan (termination is shown in Figure 3.10)

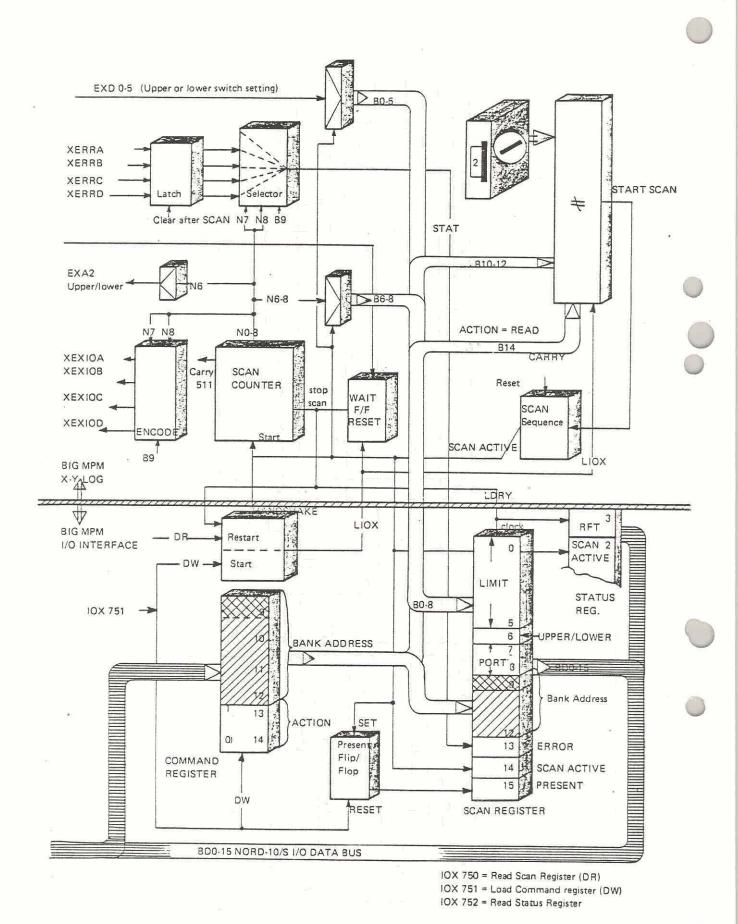


Figure 3.6: Scan of Port Status — Block Diagram

ND-06.007.01

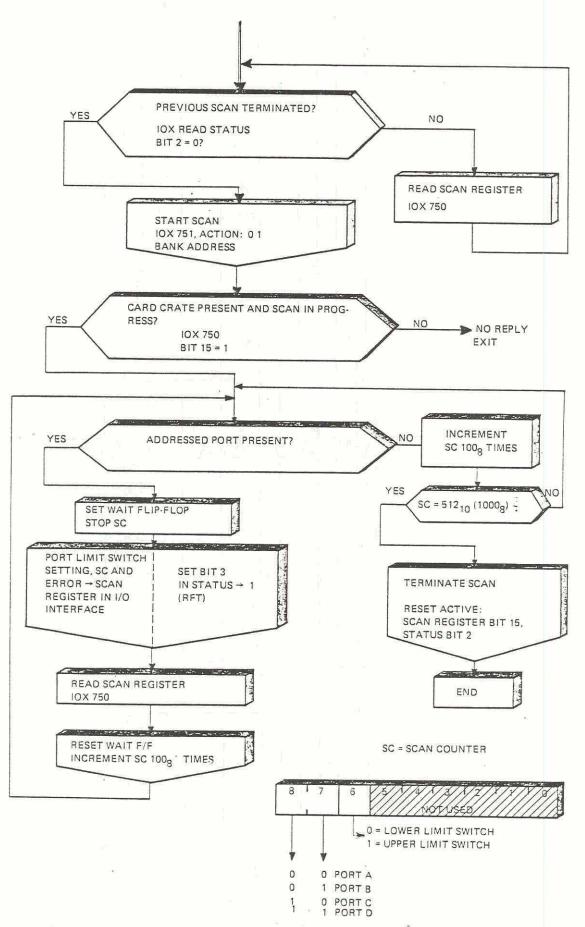


Figure 3.7: Action: 01 Scan of Port Status — Flow Diagram

ND-06.007.01

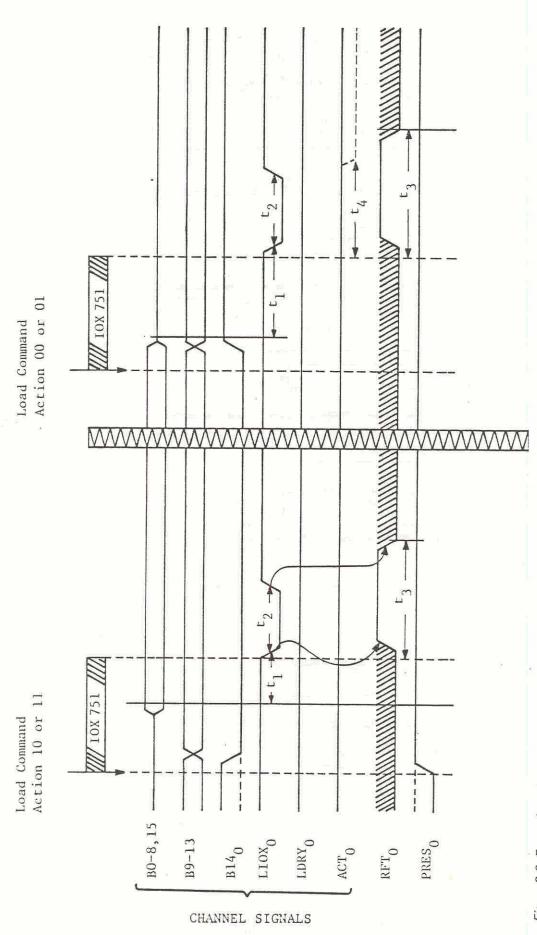


Figure 3.8: Execution of Actions 10 or 11. No answer expected from channel.

Figure 3.9: Execution of Actions 00 or 01. Addressed bank did not answer by assetting ACT with  $t_{\rm a}=400$  ns after LIOX.

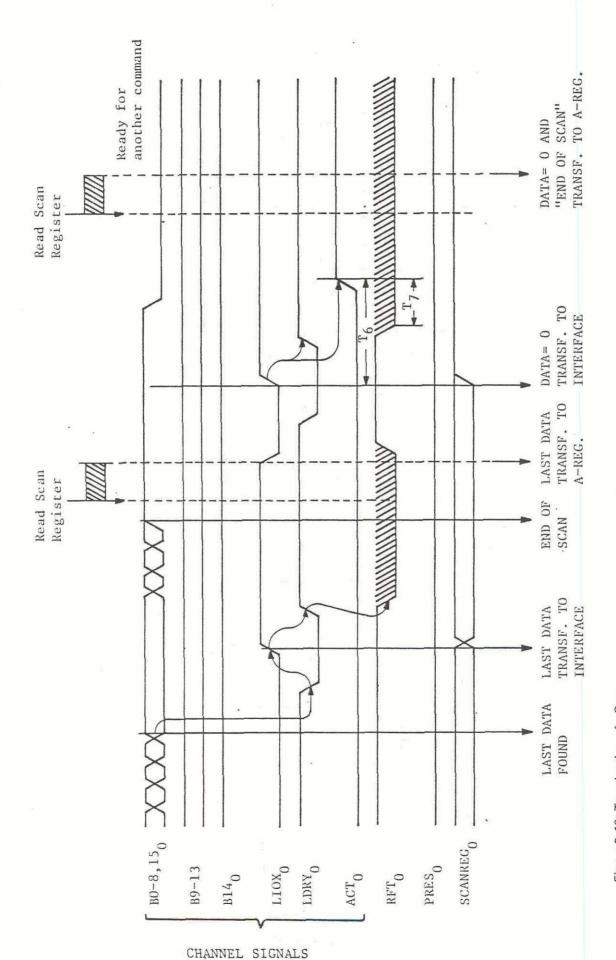


Figure 3.10: Termination of a Scan (diagram continues from Figure 3.5)

ND-06.007.01

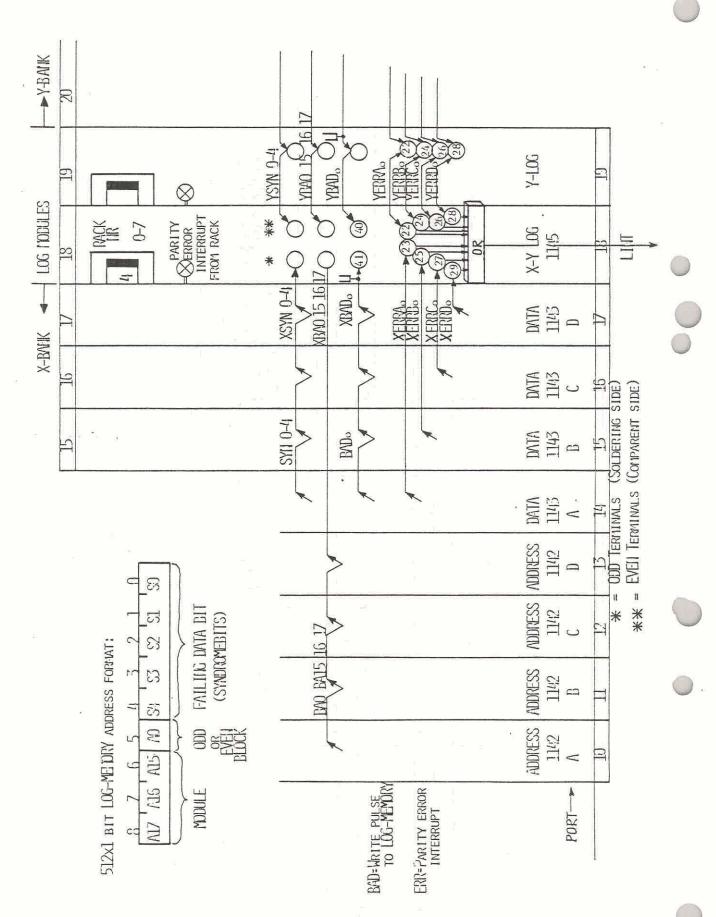
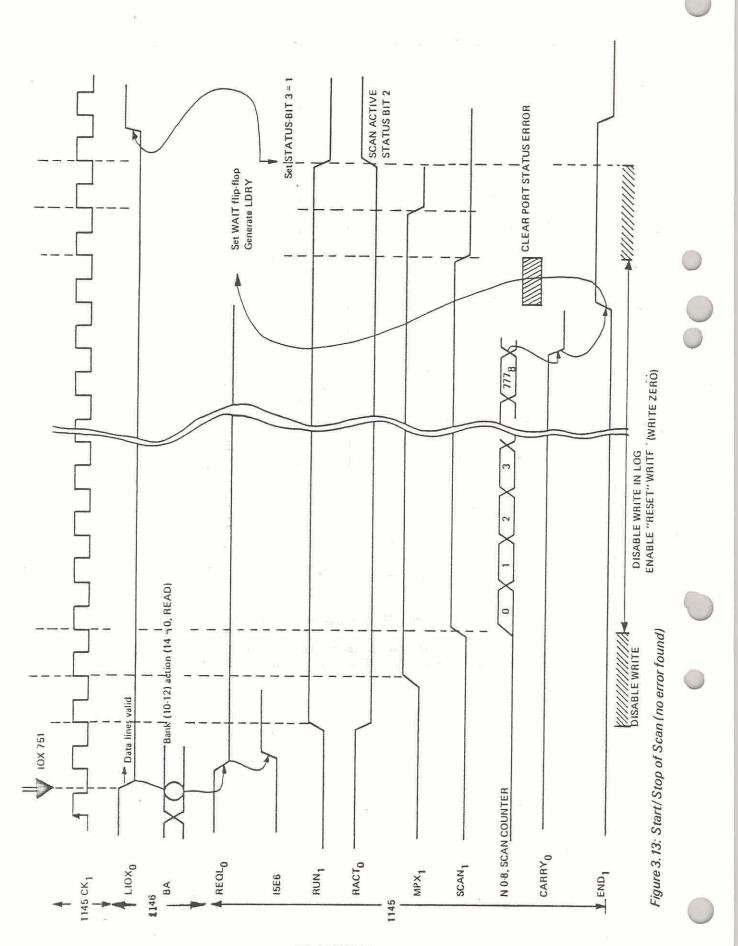


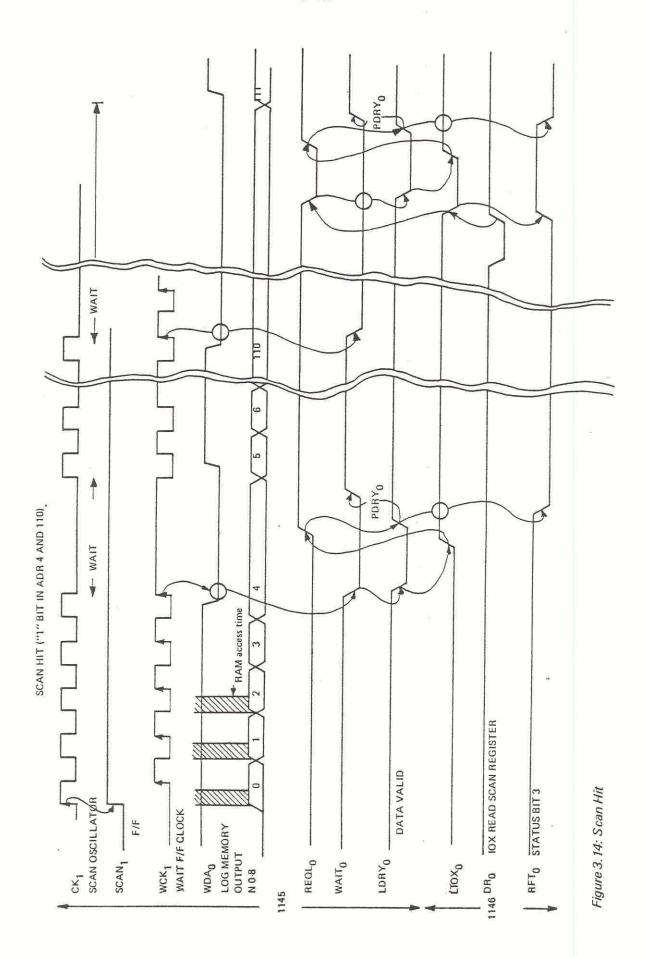
Figure 3.11: Error Log Module/Port Communication
ND-06.007.01

Description	Name	Min	Тур	Max	Unit
Data stable before LIOX	t <sub>1</sub>	200		1000	ns
Length of LIOX (scan passive)	t <sub>2</sub>	450		650	ns
Time from command to ready again (scan passive)	t <sub>3</sub>	F		700	ns
Time from LIOX to ACT if interface shall detect "crate present"	t <sub>4</sub>			400	ns
LIOX untrue after LDRY true	t <sub>5</sub>		60	100	ns
LDRY untrue to RFT true	t <sub>6</sub>		50	80	ns
Time from LIOX to ACT (exklusive cable delay)	T <sub>1</sub>		170		ns
Time from ACT to first scandata or between two scandatas	т2			85	μs
Scandata stable before LDRY	<sup>T</sup> 3	90	110		ns
LDRY true after LIOX true	т <sub>4</sub>	60		80	ns
LDRY untrue after LIOX untrue	T <sub>5</sub>		70	110	ns
ACT untrue after last LIOX untrue	т <sub>6</sub>		130	200	ns
ACT untrue after last RFT true	<sup>T</sup> 7		25	75	ns

Figure 3.12: Service Channel Timing Specifications according to Figure 3.1 - 3.4.

Times are measured at interface terminals. Times named by t are determined by interface, while times named by T are determined by  $\log$  module.





## 4 CONTROL CHANNEL SPECIFICATIONS

### 4.1 FUNCTIONAL DESCRIPTION

The purpose of the control channel is to distribute a synchronous refresh signal and 5 memory inhibits between banks.

## 4.1.1 Memory Inhibit/Power Fail

When a memory bank loses its main 5V power a memory inhibit signal, CMI, arrives from the power supply. The purpose is to inhibit all but refresh accesses to the bank. In order to maintain data while main power has disappeared, 5V, -12, and +12V stand-by power keeps vital circuitry alive.

While main power is down CMI will clear and inhibit requests from all ports. When normal power conditions return all old requests are "forgetten".

When a system is powered from different supplies a single memory channel may lose power while the bank is still going strong. Each memory channel must therefore supply a separate signal which inhibits the attached port.

When the memory channel source has common power supply to the bank, a separate inhibit signal is *not necessary*.

# 4.1.2 Refresh Synchronization

It may be advantageous to synchronize refresh in two banks which are connected to a 32 bit channel. (However, the benefits are marginal.) Via the control channel up to 8 banks may be synchronized to the same refresh oscillator. To do so one bank must be selected as a master which transmits the refresh oscillator to a number of slaves.

Master, slaves and independents are selected by backwiring straps as described in Section 5.2.2.

### 4.2 ELECTRICAL SPECIFICATIONS

Memory inhibits are single ending signals, low true, twisted with ground. Each receiver offers a standard TTL load with a pull-up and filter capacitor.

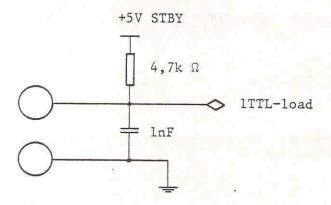


Figure 4.1: Electrical Specifications

No external termination shall be used.

The refresh oscillator is transmitted on a differntial pair from driver DS 8831 to receivers DS 8820A. Termination is only required when the signal is brought out of the cabinet of the master.

## 4.3 CHANNEL SIGNALS

The meaning of the signals are:

CMI:

Memory inhibit from the power supply of the bank.

REFOSC:

Refresh oscillator sent from a "master bank".

MIA - MID:

4 individual memory inhibits from the power supplies of

the memory channel sources A-D.

- 5 INSTALLATION
- 5.1 SWITCH SETTING
- 5.1.1 Address Range Selection

The address range has to be defined at each port of a bank. On address module 1142, two pairs of thumb wheel switches define lower and upper address limits.

Figure 5.1: Limit Switches

The switch range is 0-9 but 8 and 9 are unused (they correspond to 0 and 1 respectively). The switch setting for some limits are shown in Figure 5.2.

Be careful that the 2 address ranges do not overlap on the same channel.

If a channel shall have access to the whole bank, then

U-L = Number of 32K modules in bank.

### For Example:

With access to 0 - 256K in a bank, the limits shall be:  $U = 10_8$ , L = 0 and U-L = 8 modules.

Limit:	Address:
00	0
01	32K
02	64K
03	96K
04	128K
05	160K
06	192K
07	224K
10	256K
1	:
20	512K
30	768K
2	:
40	1024K
1	:
50	1280K
7	*
60	1536K
:	:
70	1792K
ž.	:
77	2016K

Figure 5.2: Limit Switch Setting

# 5.1.2 Crate Number Setting

In order to determine the exact position of a failing memory IC, the crates must be numbered uniquely. On the log module 1145 (or 1157) a 0-9 thumb wheel switch is mounted.

0-7 defines crate number

8-9 means crate disregarded by I/O interface

It is immaterial which number is assigned which crate, provided all are different.

#### 5.1.3 Manual Disabling of Error Correction

For maintenance purposes it may be desireable to disable error correction at a port. On all data modules 1143, a disable switch is mounted.

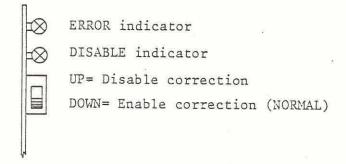


Figure 5.3: Indicators

Normally the switch shall be in the down position (correction enabled) and both indicators turned off.

Note that the disable indicator may be lighted even if the switch is turned down when correction is disabled by program (see Chapter 3).

The error indicator may be cleared by toggling the switch once.

#### 5.2 CONTROL CHANNEL CONNECTION

#### 5.2.1 Memory Inhibit

In each bank there are up to 5 separate memory inhibit signals, MI, that must be handled. Each port must receive a MI from the power supply which feeds the channel source (refer to Figure 5.4).

Channels which have power supply in common with the crate need no separate MI. Thus, in a normal single computer system only CMI is necessary.

#### 5.2.2 Refresh Synchronization

It is possible to synchronize the refresh on several banks. To do so, one bank must be selected as a *master* which transmits the refresh oscillator to a number of *slaves*. The control channel contains a pair of lines (92-93) which is intended for distribution of the refresh oscillator signal.

To participate as a master or a slave one track in the printed backwiring must be broken. Be careful to appoint only *one* master.

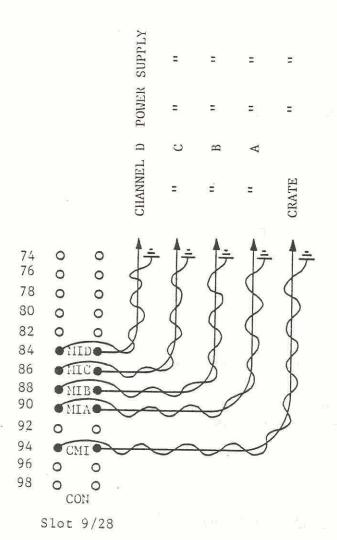


Figure 5.4: Memory Inhibits Connections in Backwiring

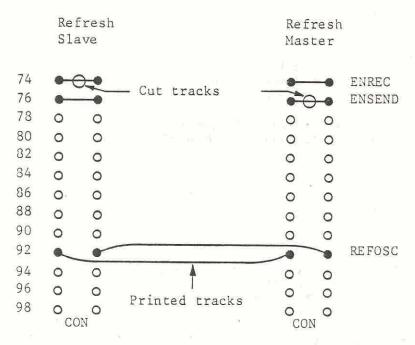


Figure 5.5: Refresh Synchronization at Control Modules in Slot 9 or 28.

Asynchronous independent bank	Do nothing (normal)
Synchronous master bank	Break at control plug 76-77
Synchronous slave bank	Break at control plug 74-75

Figure 5.6: Three Modes of Refresh Operation for a Bank

#### 5.3 CHANNEL EXPANSION

#### 5.3.1 Daisy Chaining

Linking a channel between several banks is performed by daisy chained cables. Hence, two joining plugs for entrance and exit are available for each cable.

Note! All channels between two banks in one crate are internally daisy chained.

Figurer 5.7 shows which plug fields are internally connected.

Entrance and exit may be exchanged since all connections are 1:1.

	Slots:	Terminal Range:
Port A Address	10 - 27	95 - 50
Port A Data	14 - 23	95 - 56
Port B Address	11 - 26	95 - 50
Port B Data	15 - 22	95 - 56
Port C Address	12 - 25	95 - 50
Port C Data	16 - 21	95 - 56
Port D Address	13 - 24	95 - 50
Port D Data	17 - 20	95 - 56
Maintenance Channel	18 - 19	95 - 56
Control Channel	9 - 28	95 - 84

Figure 5.7: Internally Connected Plug Fields Used for Channel Expansion

It is strictly forbidden to carry out other signals than those stated in "terminal range" in Figure 5.7.

#### For Example:

If 25 pairs of flat cables are used the forbidden signal wires must be properly broken.

#### 5.3.2 Termination

All systems with channel cable length of more than 1 meter (or between cabinets) shall be terminated. Thus, single bank systems (ND-143, ND-144) contained in one cabinet do not need termination but all others do. The terminal plug is put on the last free contract at the end of the channel. Except for the control channel all signal pairs shall be terminated in 120  $\Omega$ . On the control channel only the REF OSC signal (pair 92 - 93) shall be terminated if synchronized refresh is used and brought out of the cabinet.

#### 5.4 CONFIGURATION STRUCTURE

There are 4 basic bank configurations: ND-143, 144, 146 and 158 which may be expanded by ports ND-145, 147 and storage modules ND-156.

The modules that constitute the various ND numbers are shown in Appendix F. This appendix also shows the proper slot numbers for each module. The primary positions are shown with a bold frame. Expansion possibilities are indicated by a light frame.

Note! ND-146 and 158 require extra wire-wrap backwiring unless a separate flat cable plug panel is used.

#### APPENDIX A

#### BMPM TO MPM COMPARISON

Big multiport memory (BMPM) compared to the former model of multiport memory (MPM). Below are short statements regarding the differences and similarities between the two models:

- 1. Same concept and design philosophy.
- Improved electrical specifications allow 10 ports per channel on BMPM compared to 4 ports on MPM.
- 3. Signal definitions and timing requirements of the channel are the same.
- 4. BMPM access time are somewhat faster.
- 5. BMPM ports may be connected to existing MPM channels.
- 6. One port serves one bank in BMPM while it serves two in MPM.
- 7. BMPM address range is 2048K while it is 256K in MPM.
- 8. Resolution of port address limits are 32K on BMPM while 8K on MPM.
- Bank sizes, in one crate, are 256K and 128K in BMPM compared to 2 x 64K in MPM.
- BMPM has error correction as standard (stored data has 16 + 5 bits) while MPM has dual parity error detection (stored data has 16 + 2 bits).
- 8K memory modules (1094) ND-152 cannot be used in BMPM. 32K memory modules (1132) ND-156 can be used in MPM but requires modification of refresh (ECO issued).
- In contrast to MPM, BMPM features a special maintenance channel for reporting of errors, reading of address range switches and general test purposes.

# APPENDIX B CONVERSION CHARTS

## B.1 CONVERSION BETWEEN LOGICAL CODE AND PHYSICAL POSITION OF FAULTY IC'S ON 1132 MODULE

IC CODE	1132	1	
	COORDINATES	CODE	1132 COORDINATES
0 1 2	- No error 1 0 B 17 C	40 41 42	- No error 10 D 17 E
3 4 5 6	3 B 6 B 13 C - Mult. error	43 44 45 46	3 D 6 D 13 E - Mult. error
7 10 11	in B or C 4 C 1 C 10 C - Mult. error	47 50 51	in D or E 4 E 1 E 10 E
12 13 14	in B or C 6 C	52 53 54	- Mult. error in D or E 6 E 4 D
15 16 17	9 B 3 C - Mult. error in B or C	55 56 57	9 D 3 E - Mult. error in D or E
20 21 22 23	9 C 19 C 14 B 11 B	60 61 62 63	9 E 19 E 14 D
24 25 26 27	7 B - Mult. error in B or C 7 C - Mult. error in B og C	64 65 66	7 D - Mult. error in D or E 7 E - Mult. error
30 31 32 33	14 C - Mult. error in B or C 1 3 B - Mult. error	70 71 72	in D or E  14 E  - Mult. error  in D or E  13 D  - Mult. error
34 35 36 37	in B or C 16 C - Mult. error in B or C - Mult. error in B or C - Mult. error	73 74 75 76	in D or E 16 E - Mult. error in D or E - Mult. error in D or E - Mult. error

B.2 CONVERSION BETWEEN LOGICAL CODE AND PHYSICAL SLOT OF FAULTY STORAGE MODULES

MODULE CODE	CRATE SLOT	MODULE CODE	CRATE SLOT
0	1	10	32
1	2	11	31
2	3	12	30
3	4	13	29
4	5	14	
5	6	15	
6	7	16	Unassigned
7	8	17	ال ا

### B.3 PHYSICAL POSITION OF RANGE LIMIT SWITCHES INDICATED BY THE BANK/PORT CODE

BANK/PORT CODE 987	PORT	CRATE SLOT
000	XA	10
001	XB	12
010	XC	.14
011	XD	16
100	YA	27
101	YB	25
110	YC	23
111	YD	21

### APPENDIX C

#### CABLE AND WIRING LISTS

Figure C.1: Memory Channel — Address

Figure C.2: Memory Channel — Data

Figure C.3: Service Channel

Figure C.4: Control Channel

Figure C.5: Wire Wrapping on ND-146

Figure C.6: Terminal Plug

Figure C.7: Memory Channel — Address Interleave Cable

	IORSK DATA- EKTRONIKK	BMP Add	M Memory Channe resses at Port	1, produce		I
. NO.	SIGNAL	POLARITY	ADDRESS PORT 1:1 cable	ADDRESS PORT Interleave cable	a H	
1	BA15L BA15L	0	BERG 95	93		
	BA14L	1 0	" 94	92 91 .		
2	BA14L	1	" 92	90		
3	BA13L	0	" 91 " 00	89		
	BA13L BA12L	0	" 90 " 89	88		
4	BA12L	1	11 88	86		<del> </del>
5	BALLL	0	" 87	85	***	
	BAllL	1	" 86	84		
6	BA10L BA10L	0	" 85 " 84	83	************	
	BA9L	0	" 83	81	•	
7	BA9L	1	" 82	80		
8	BA8L	0	" 81	. 79		
~\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\	- BASL	1	" 80 " 79	78		
9	BA7L BA7L	0 1	" 78	77		
10	BA6L	ō	0 77	75		
10	BA6L	1	" 76	74		
11	BA5L BA5L	0	11 75	73		
	BA4L	1 0	" 73	72 71		
12	BA4L	1	72	70		
13	BA3L	0	71	69		
	BA3L	1	" 70	68		
14	BA2L BA2L	0	" 69 " 68	67		
	BAIL	0	" 67	65		
15	BA1L	1	" 66	64		
16	BAOL	0	11 65	51		
	BAOL	1	11 64	50		
17	BA17L BA17L	0	" 63 " 62	61		
1.0	BA16L	0	" 61	95		
18	BA16L	1	60	94		
19	WRITEL	0	" 59	59		
	WRITEL REQL	1 1	58	58		
20	REQL	0	" 57 " -56	57 56		
24	BA18	0		63		
21	BA18	1	" 54	62	-	
22	BA19	0	" 53	55		
	BA19	1 1	" 52 " 51	54		
23	BA20 BA20	0	" 51 " 50	53	-	
		Sta	with the same of t	ential channel	7	1
RAWN	ВУ	Remarks	arrier c	ar charme		net - Det-
			91 = 161			nt for Date
PROVE	D BY	Sig	mals 1-20 are	Standard Extended Addres	Replaced b	y Date

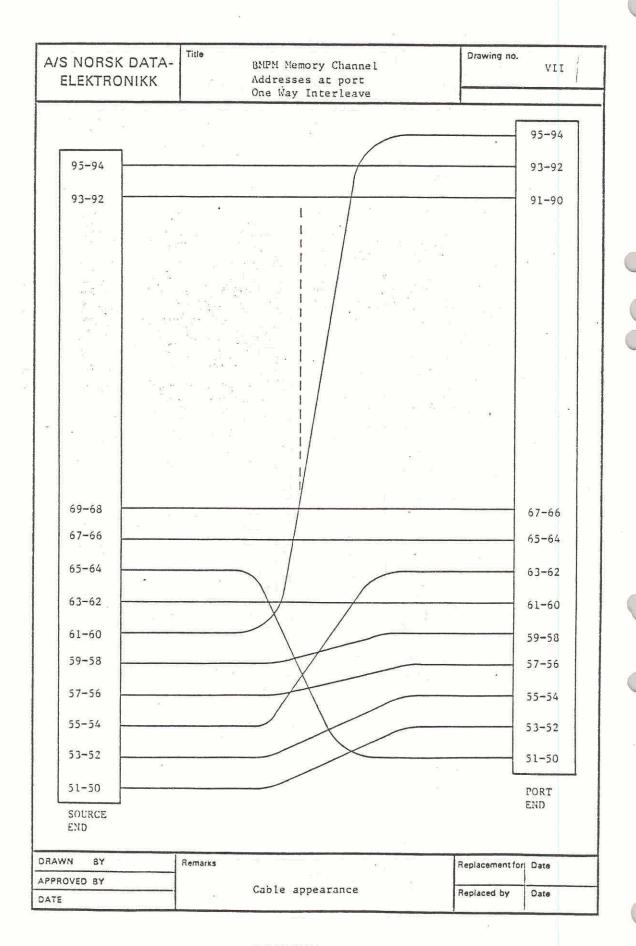
	ORSK DATA- KTRONIKK	Tide E	SMPM Memory Channe Data at Port	1	Drawing no.	
NO.	SIGNAL	POLARITY	DATA PORT			
1	BD15L	0	BERG 95			
	BD15L BD14L	0	94			
2	BD14L	1	92			
3	BD13L	0	" 91			
	BD13L BD12L	1	" 90			
4	BD12L	0	" 89 " 88			
5	BD11L	0	" 87			
	BD11L	1	" 86			
6	BD10L BD10L	0	" 85 " 84			
	BD9L	0	" 83		· · · · · · · · · · · · · · · · · · ·	
. 7	BD9L	1	" 82			
8 -	BDSL	0	81			
-	BD7L	1 0	" 80 " 79			
9	BD7L	1	78			
10	BD6L	0	" 77			
	BD6L BD5L	0	76			
11	BD5L	1	75			
12	BD4L	0	" 73			
	BD4L	1	72			
13	BD3L BD3L	0	<del>                                     </del>			
14	BD2L	0	70 69			
14	BD2L	1	" 68			
1.5	BD1L	0	6/			
1.6	BDIL	0	66			
16	BDOL	1	64			
17	BD17L	0	" 63			
	BD17L BD16L	0	62			
18	BD16L	1	" 60			
19	ARL	. 0	" 59			
	ARL DRL	0	58			
20	DRL	1	" 56_			
20	DRL Standai	1		i i	- S	
DRAWN	ev 7	P				
	BY	Remarks		la la	Replacement for Date	
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STAC			× = = = = = = = = = = = = = = = = = = =	e <sub>g</sub> a es <sup>2</sup>		

	RSK DATA- TRONIKK	Title .	MPM Service Channel		Drawing no.
NO.	SIGNAL	POLARITY	PLUG BERG (CPU POS.)		
1	B15L	0 -	BERG 95		
-	BISL B14L	0	" 94		
2	B14L	1	92		
3	B13L	0	'' 91 '' 90		
	B13L B12L	0	" 90 " 89		
4	B12L	1	" 88		
5	BliL	0	" 87 " 86		
	B11L B10L	0	" 86 " 85		
6	BlOL	1	" 84 -		
7	B9L	0	" 33		
	B9L B8L	1 0	" 82 " 31		
- 8-	B8L	1	" 80		
9	B7L	0	79		
	B7L B6L	0	78		
10	B6L	1	" 76		
11	B5L	0	75		
	B5L B4L	0	" 74 " 73		
12	B4L	1	11 72		
13	B3L	0	71		
	B3L B2L	0	" 70 " 69		
14	B2L	1	" 68		
15	BIL	0	11 67		
	B1L B0L	1 0	" 66 " 65		
16	BOL	1	11 64		
17	LINTL	0	11 63		
1.0	LINTL	0	62		
18	LDRYL	1	" 60		
19	ACTL ACTL	0-	11 59		
	LIOXL	0	" 58 " 57		
20	LIOXL	1	" 56.		
	THE REAL PROPERTY AND ADDRESS OF THE PERSON NAMED IN COLUMN TWO IN COLUM		ential channel		
DRAWN	вү	Romarks			Replacement for Date
PPROVED	ву	3801	X		
ATE				****	Replaced by Date

II.	RSK DATA- TRONIKK		PM Control Ch	annel	Drawing no	IV
NO.	SIGNAL	POLARITY	PLUG BERG (CPU POS.)			
1	CMI CMD	0	BERG 95			
2	REFOSC	0	93			
3	REFOSC MIA	0	" 92 " 91			
	GND MIB	0	" 90 " 89			
4	GND MIC	14,0	" 88			
5	GND	0	" 86			
6	MID GND	0	" 85 " 84			
7		-	41 11			
8 _		i	11			
9	*	343	n v			
10	ENSEND	0	77	, Connected		
	GND ENREC	1 0	" 76 " 75	in backwir.	STRAPS FOR SYNCHRONIZ	REFRESH
11	GND'	1 0	74	in backwir.	- JINCHROMIZ	ING
12		1	11			
13		0 1	u L			
14		0	и			
15		0	0			
16		0	-11			
17		1 0	U			
		1 0	11			
18		1 0	11			
19		1	<u> </u>			
20		0	11			
	).		:# =	:	24	
			- ×		2 2012 2012	
DRAWN	EY	Romarks		."	Replacement	for Date
APPROVED	ВУ		- E	•a _a	Replaced by	Date
DATE		-				

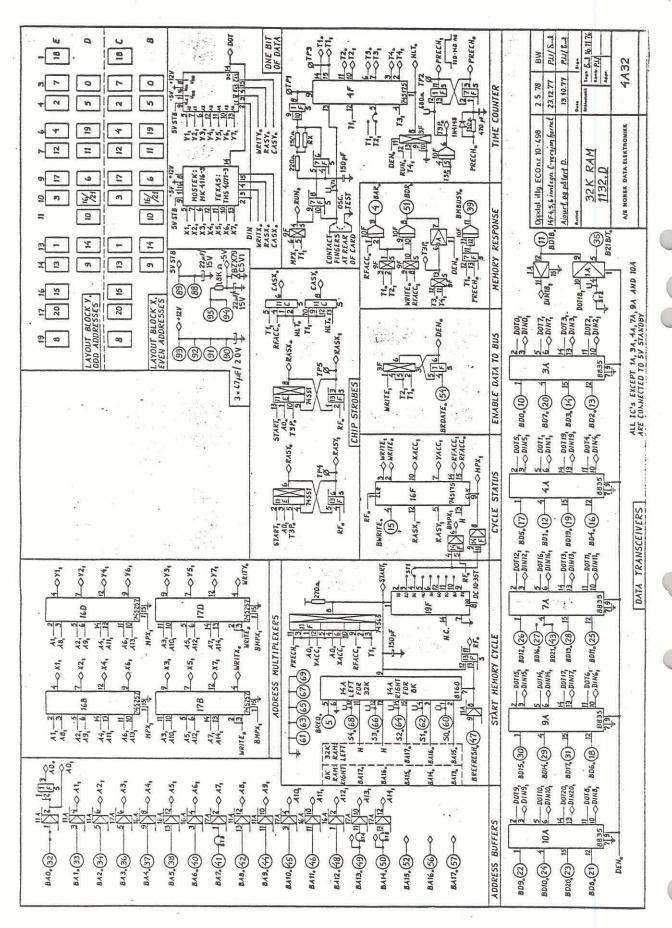
A/S NORSK DATA- ELEKTRONIKK	BMPM Wire Won ND-	Drawing no.	
*			
XDDAT - YDDAT	'XCDAT - YCDAT	XBDAT - YBDAT	XADAT - YADAT
17:56 - 20:56	16:56 - 21:56	15:56 - 22:56	14:56 - 23:56
17:57 - 20:57	1 1		
Í	- 1		1
1			i
17:94 - 20:94	1	1	I I
17:95 - 20:95	16:95 - 21:95	15:95 - 22:95	14:95 - 23:95
- XDADR - YDADR	XCADR - YCADR	XBADR - YBADR	XAADR - YAADR
13:56 - 24:56	12:56 - 25:56	11:56 - 26:56	10:56 - 27:56
- 1		T.	1
l I			1
1	1 -		!
- 1			1
13:95 - 24:95	12:95 - 25:95	11:95 - 26:95	10:95 - 27:95
_			
In all 8 groups	40 wires are wrappe		of same number.
			8
YE NWARD	Remarks		Replacement for Date
APPROVED BY			

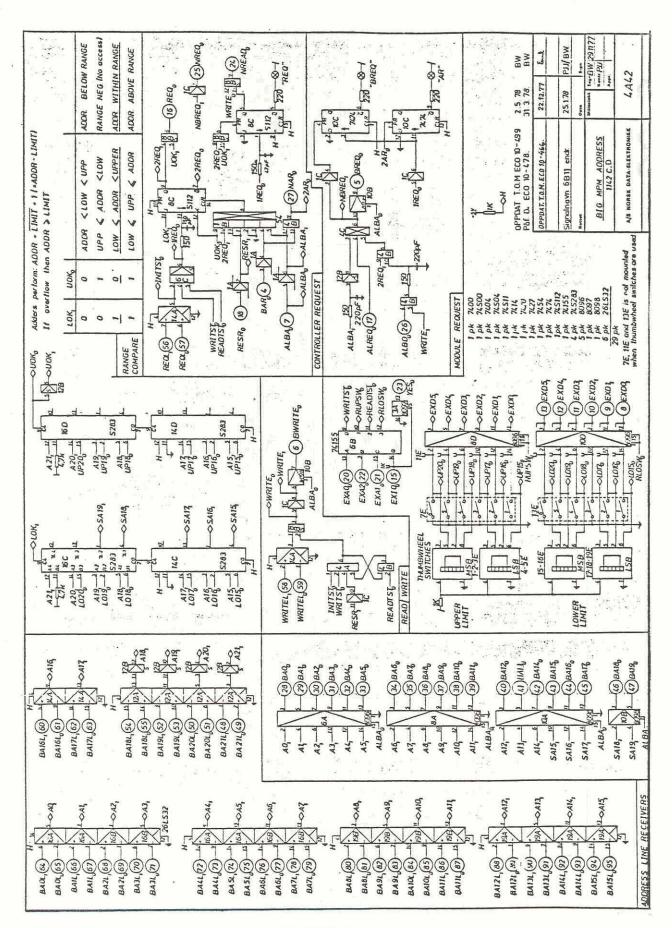
A/S NORSK DATA- ELEKTRONIKK	Title BMPM Termination Plugs	Drawing no.	VI i
* Memory Char	rs, 120 Ω mel Address  Memory Channel Data Maintenance Channel  Te 2 x 26 pins BERG 65051-011	resistor,	120 Ω nel
	E Z X ZO PING SENG GOODI-UII		
10 to			
DRAWN BY APPROVED BY DATE		Replacement for I	Date

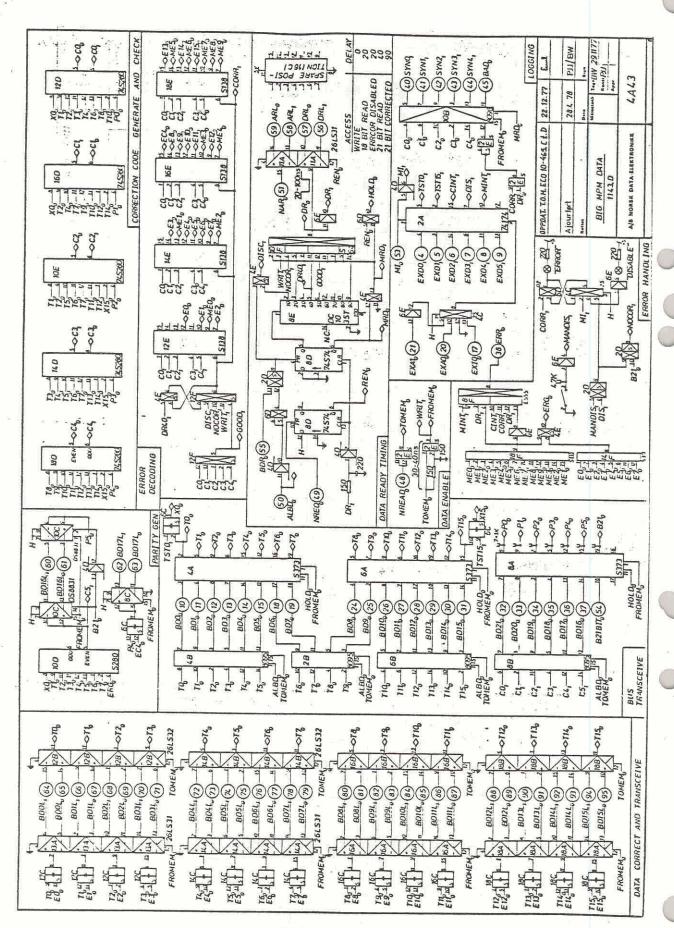


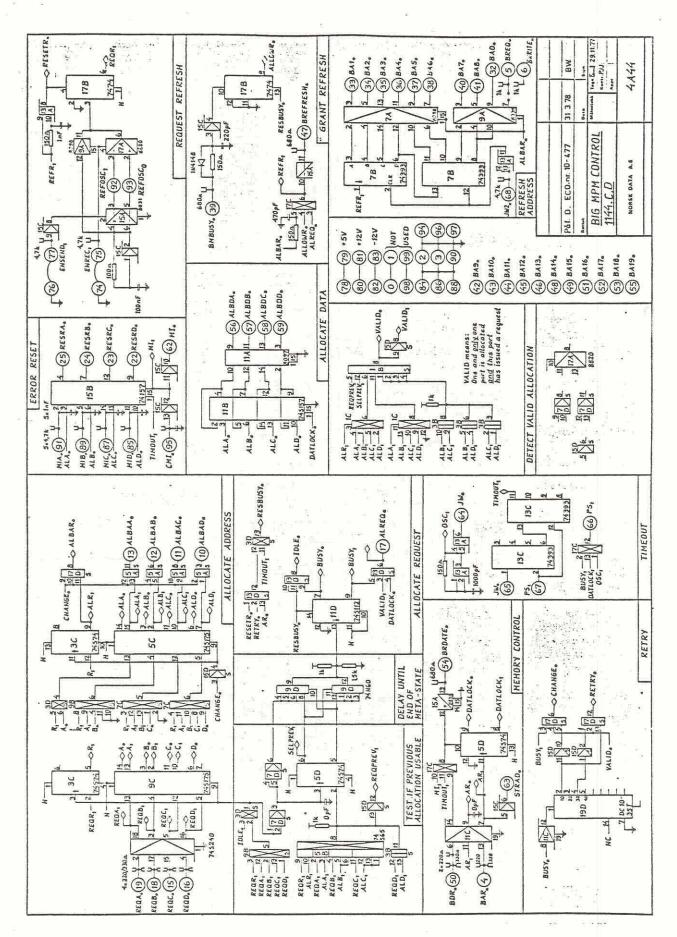
APPENDIX D

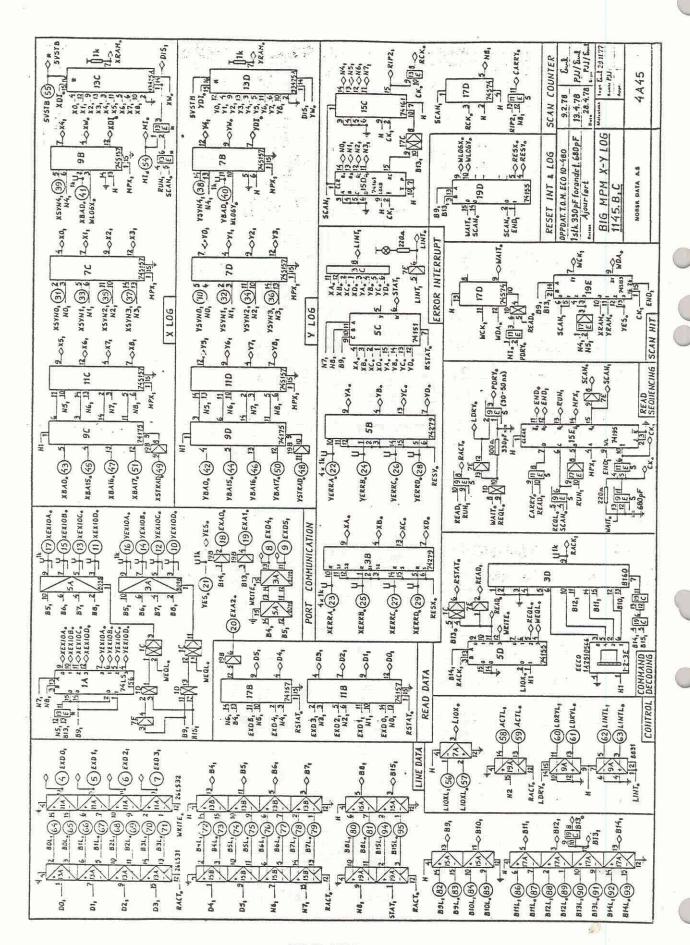
LOGIC DIAGRAMS

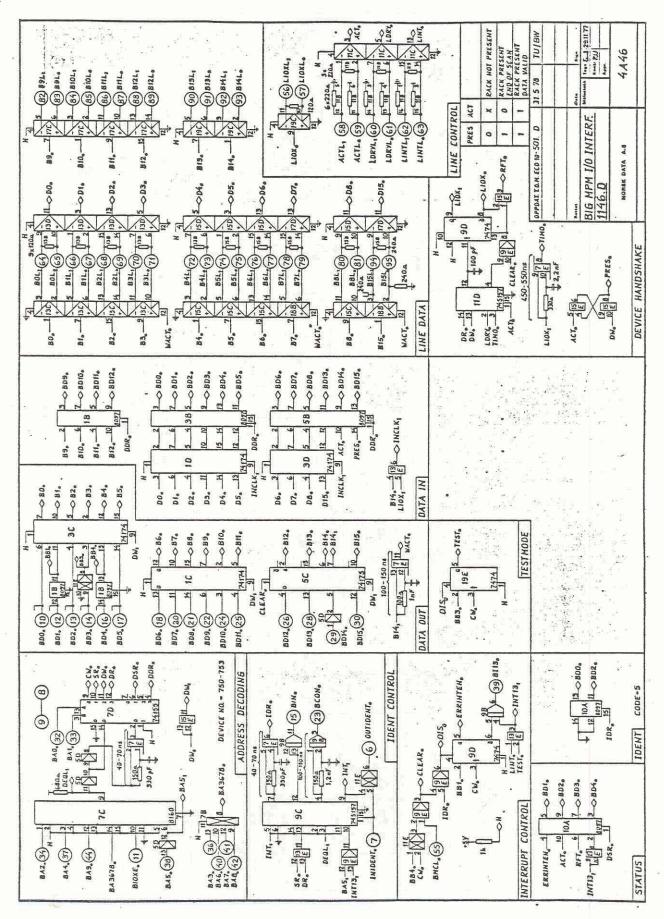


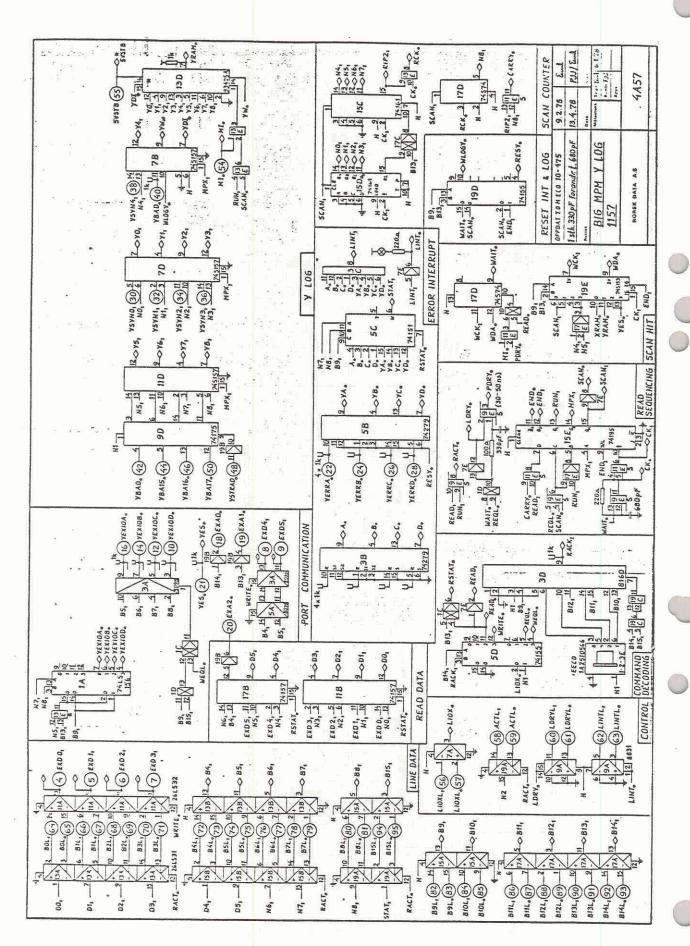












#### APPENDIX E

### RS-422 STANDARDS

### E.1 SUMMARY OF EIA RS-422 STANDARD FOR A BALANCED DIFFERENTIAL INTERFACE

A.	Line Driver	8.
	Open Circuit Voltage (either logic state) Differential  V <sub>do</sub>   = 6.0V Common Mode  V <sub>cmol</sub> = 3.0V	
	Differential Output Voltage (across 100 ohm load)  Either logic state $ V_d  = max (0.5V_{dec} 2.0V)$	
	Output Impedance Either logic state R <sub>G</sub> ~ 100 ohms	
	$\begin{array}{lll} \text{Mark-Space Level Symmetry (across 100 ohm load)} \\ \text{Differential} &  V_{dS}  -  V_{dM}  \approx 0.4V \\ \text{Common Mode} &  V_{cmS}  -  V_{cmM}  \approx 0.4V \end{array}$	
	Output Short Circuit Current (to ground)  Either Output  I <sub>SC</sub>   < 150mA	
	Output Leakage Current (power off)  Voltage Range $-0.25V \le V_x \le +6.0V$ Either Output at $V_x$ $ I_x  \le 100 \mu A$	
	Rise and Fall Times (across 100 ohm load)	
	$T = Baud Interval$ $(t_r, t_l) = max (0.1T, 20ns)$	
	Ringing (across 100 ohm load)  Definitions  V <sub>dSS</sub> = V <sub>d</sub> (steady state)  V <sub>SS</sub> = V <sub>dS</sub> - V <sub>dM</sub> (steady state)	
	Limits (either logic state)	
	Percentage $ V_d - V_{dSS}  \le 0.1V_{SS}$ Absolute $2.0V -  V_d  \le 6.0V$	
		1
		11

8. 1	Line Receiver	
5	Signal Voltage Range	
	Differential	V <sub>d</sub>   ≤ 6.0\
	Common Mode	IV = 1.01
5	Single-Ended Input Current (power Of	N or OFF)
	Either Input at V <sub>x</sub>	Vx  = 10
	Other Input Grounded	ly  = 3.25m/
5	Single-Ended Input Bias Voltage (othe	r input grounded)
	Either Input Open Circuit	Va  = 3.01
5	Single-Ended Input Impedance (other	input grounded)
	Either Input	RL = 4000 ohms
1	Differential Threshold Sensitivity	
	Common Mode Voltage Range	V <sub>cm</sub>   ≤ 7.0\
	Either Logic State	VT = 200m
1	Absolute Maximum Input Voltage	
	Differential	Va  = 12\
1	Single-Ended	Vx   = 104
1	nput Balance (threshold shift)	
	Common Mode Voltage Range	V <sub>cm</sub>   ≤ 7.0\
	Differential Threshold (500 ohms in	series with each
i	nput)	
	Either Logic State	$ V_t  \approx 400 \text{mV}$
T	ermination (optional)	No.
	Total Load Resistance (differential)	$R_{\tau} > 90$ ohms
٨	Aultiple Receivers (bus applications)	
	Up to 10 receivers allowed. Different	
60	sitivity of 200mV must be maintaine	a.
H	lysteresis (optional)	180 STATE OF STATE
	As required for applications with slo receiver, to control oscillations.	w rise/fall time at
F	ail Safe (optional)	*
	As required by application to provide	a steady MARK or
	SPACE condition under open con	pector or driver

As required by application to provide a steady MARK or SPACE condition under open connector or driver ower

OFF condition.

#### C. Interconnecting Cable

Туре

Twisted Pair Wire or Flat Cable Conductor Pair

Conductor Size

Copper Wire (solid or stranded) Other (per conductor) 24 AWG or larger R ≤ 30 ohms/1000 ft.

Capacitance

Mutual Pair Stray C = 20pFift. C = 40pFift.

Pair-to-Pair Cross Talk (balanced)

Attenuation at 150KHz

A = 40dB

#### E.2 EIA STANDARD RS-422

Recommendation V.11

# ELECTRICAL CHARACTERISTICS FOR BALANCED DOUBLE-CURRENT INTERCHANGE CIRCUITS FOR GENERAL USE WITH INTEGRATED CIRCUIT EQUIPMENT IN THE FIELD OF DATA COMMUNICATIONS 9

(Geneva, 1976)

#### 1. Introduction

This Recommendation deals with the electrical characteristics of the generator, receiver and interconnecting leads of a differential signalling (balanced) interchange circuit with an optional d.c. offset.

The balanced generator and load components are designed to cause minimum mutual interference with adjacent balanced or unbalanced interchange circuits (see Recommendation V.10) provided that waveshaping is employed on the unbalanced circuits.

In the context of this Recommendation, a balanced interchange circuit is defined as consisting of a balanced generator connected by a balanced interconnecting pair to a balanced receiver. For a balanced generator the algebraic sum of both the outlet potentials, with respect to earth, shall be constant for all signals transmitted; the impedances of the outlets with respect to earth shall be equal. The degree of balance of the interconnecting pair is a matter for further study.

Annexes are provided to give guidance on a number of application aspects as follows:

Annex 1 Cables and terminations

Annex 2 Compatibility with other interfaces

Annex 3 Multipoint operation

Note. - Generator and load devices meeting the electrical characteristics of this Recommendation need not operate over the entire signalling rate range specified. They may be designed to operate over narrower ranges to satisfy requirements more economically, particularly at lower signalling rates.

Reference measurements are described which may be used to verify certain of the recommended parameters but it is a matter for individual manufacturers to decide what tests are necessary to ensure compliance with the Recommendation.

#### Field of application

The electrical characteristics specified in this Recommendation apply to interchange circuits operating with data signalling rates up to 10 Mbit/s, and are intended to be used primarily in Data Terminal Equipment (DTE) and Data Circuit-terminating Equipment (DCE) implemented in integrated-circuit technology.

This Recommendation applies to new work and is not intended to apply to DCE implemented in discrete component technology, for which the electrical characteristics covered by Recommendation V.28 are more appropriate.

<sup>5)</sup> This Recommendation is also designated as X.27 in the Series X Recommendations.

Typical points of application are illustrated in Figure 1/V.11.

Whilst the balanced interchange circuit is primarily intended for use at the higher signalling rates, its use at the lower rates may be necessary in the following cases:

- 1) where the interconnecting cable is too long for proper unbalanced circuit operation;
- 2) where extraneous noise sources make unbalanced circuit operation impossible;
- 3) where it is necessary to minimize interference with other signals.

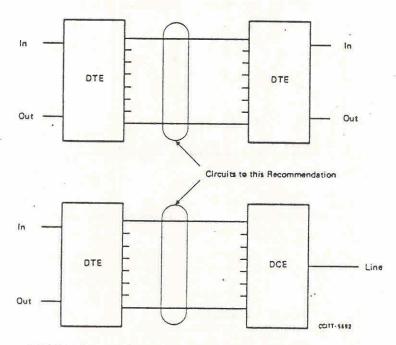
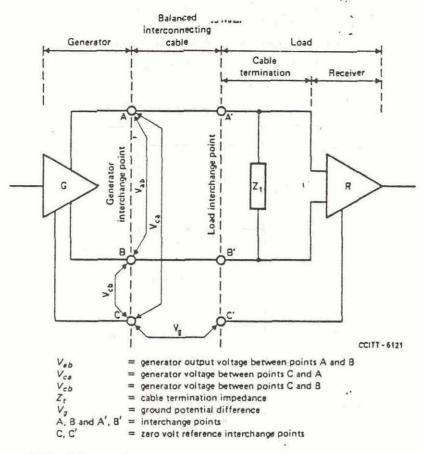


FIGURE 1/V.11 - Typical applications of balanced interchange circuits

#### 3. Symbolic representation of interchange circuit (Figure 2/V.11)



FIGURF 2/V.11 - Symbolic representation of a balanced interchange circuit

Note 1. — Two interchange points are shown above. The output characteristics of the generator, excluding any interconnecting cable, are defined at the "generator interchange point". The electrical characteristics to which the receiver must respond are defined at the "load interchange point".

Note 2. - Point C may be connected to C' by CCITT Recommendation V.24 Circuit 102 and to protective ground if required by national regulations.

#### 4. Generator polarities and receiver significant levels

#### 4.1 Generator

The signal conditions for the generator are specified in terms of the voltage between output points A and B shown in Figure 2/V.11.

When the signal condition 0 (space) for data circuits or ON for control and timing circuits is transmitted, the output point A is positive with respect to point B. When the signal condition 1 (mark) for data circuits or OH for control and timing circuits is transmitted, the output point A is negative with respect to point B.

#### 4.2 Receiver

The receiver differential significant levels are shown in Table A/V.11 below, where  $V_{A'}$  and  $V_{B'}$  are respectively the voltages at points A' and B' relative to point C':

TABLE A/V.11 - Receiver differential significant levels

•	$V_{A'} - V_{B'} < -0.3 \text{ V}$	$V_{A'} - V_{B'} > +0.3 \text{ V}$
Data circuits	1	0
Control and timing circuits	OFF	ON

#### Generator

#### 5.1 Resistance and offset voltage

- 1. The total generator resistance between points A and B shall be equal to or less than 100 ohms and adequately balanced with respect to point C. (It is left for further study as to the degree of balance required both statically and dynamically.)
- 2. The magnitude of the generator d.c. offset voltage (see 5.2.2 below) shall not exceed 3 V under all operating conditions.

#### 5.2 Static reference measurements

The generator characteristics are specified in accordance with measurements illustrated in Figure 3/ V.11 and described in 5.2.1 to 5.2.4 below.

#### 5.2.1 Open-circuit measurement (Figure 3a)/V.11)

The open-circuit voltage measurements are made with a 3900 ohm resistor connected between points A and B. For either binary state, the magnitude of the differential voltage  $(V_0)$  shall not be more than 6.0 volts, nor shall the magnitude of  $V_{00}$  and  $V_{00}$  be more than 6.0 volts.

#### 5.2.2 Test-termination measurement (Figure 3b)/V.11)

With a test load of two resistors, each 50 ohms, connected in series between the output points A and B, the differential voltage  $(V_i)$  shall not be less than 2.0 volts or 50% of the magnitude of  $V_0$ , whichever is greater. For the opposite binary state the polarity of  $V_i$  shall be reversed  $(V_i)$ . The difference in the magnitudes of  $V_i$  and  $V_i$  shall be less than 0.4 volts. The magnitude of the generator offset voltage  $V_0$ , measured between the centre of the test load and point C shall not be greater than 3.0 volts. The magnitude of the difference in the values of  $V_{0s}$  for one binary state and the opposite binary state shall be less than 0.4 volts.

Note. — Under some conditions this measurement does not determine the degree of balance of the internal generator impedances to point C. It is left for further study whether additional measurements are necessary to ensure adequate balance in generator output impedances.

#### 5.2.3 Short-circuit measurement (Figure 3c)/V.11)

With the output points A and B short-circuited to point C, the currents flowing through each of output points A or B shall not exceed 150 milliamperes for either logical condition.

#### 5.2.4 Power-off measurements (Figure 3d)/V.11)

Under power-off condition with voltages ranging between +0.25 volt and -0.25 volt applied between each output point and point C, as indicated in Figure 3d)/V.11, the magnitude of the output leakage currents  $(I_{xa} \text{ and } I_{xb})$  shall not exceed 100 microamperes.

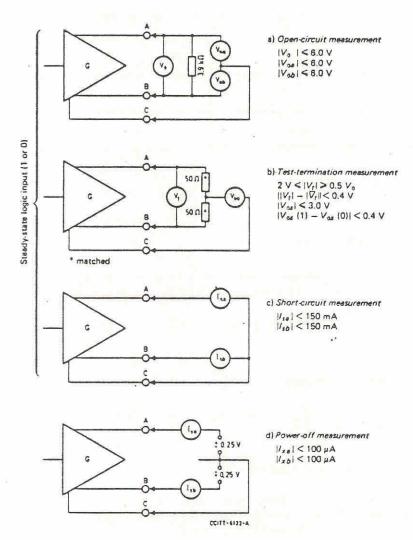
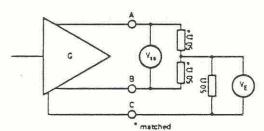


FIGURE 3/V.11 - Generator-parameter reference measurements

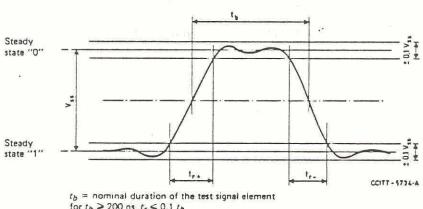
#### 5.3 Dynamic voltage balance and rise time (Figure 4/V.11)

With the measurement configuration shown in Figure 4/V.11, a test signal with a nominal signal element duration  $t_h$  and composed of alternate ones and zeros, shall be applied to the input. The change in amplitude of the output signal during transitions from one binary state to the other shall be monotonic between 0.1 and 0.9  $V_{\alpha}$  within 0.1 of  $t_h$  or 20 nanoseconds, whichever is greater. Thereafter the signal voltage shall not vary more than 10% of  $V_{\alpha}$  from the steady state value.

The resultant voltage due to imbalance  $(V_E)$  shall not exceed 0.4 V peak-to-peak (the value of  $V_E$  is provisional and is subject to further study to determine whether voltage peaks of very short duration should be included).



 $V_{\rm E} <$  0.4 V peak-to-peak (provisional)  $V_{zz} =$  difference between signal steady-state voltages



for  $t_b > 200$  ns,  $t_r < 0.1$   $t_b$  for  $t_b < 200$  ns,  $t_r < 20$  ns

FIGURF 4/V.11 - Generator dynamic balance and rise time measurement

6. Load

#### 6.1 Characteristics

The load consists of a receiver (R) and an optional cable termination resistance  $(Z_i)$  as shown in Figure 2/V.11. The electrical characteristics of the receiver are specified in terms of the measurements illustrated in Figures 5/V.11, 6/V.11 and 7/V.11 and described in 6.2, 6.3 and 6.4 below. A circuit meeting these requirements results in a differential receiver having a high input impedance, a small input threshold transition region between -0.3 and +0.3 volts differential, and allowance for an internal bias voltage not to exceed 3 volts in magnitude. The receiver is electrically identical to that specified for the unbalanced receiver in Recommendation V.10.

#### 6.2 Receiver input voltage - current measurements (Figure 5/V.11)

With the voltage  $V_{i,i}$  (or  $V_{i,h}$ ) ranging between -10 volts and +10 volts, while  $V_{i,h}$  (or  $V_{i,h}$ ) is held at 0 volt, the resultant input current  $I_{i,a}$  (or  $I_{i,h}$ ) shall remain within the shaded range shown in Figure 5/V.11. These measurements apply with the power supply of the receiver in both the power-on and power-off conditions.

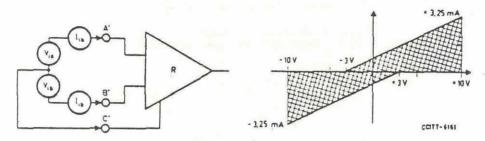


FIGURE 5/V.11 - Receiver input voltage-current measurements

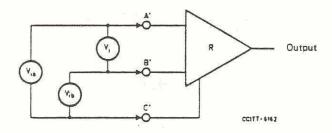
#### 6.3 D.c. input sensitivity measurements (Figure 6/V.11)

Over the entire common mode voltage  $(V_{cm})$  range of +7 volts to -7 volts, the receiver shall not require a differential input voltage  $(V_i)$  of more than 300 millivolts to assume correctly the intended binary state. Reversing the polarity of  $V_i$  shall cause the receiver to assume the opposite binary state.

The maximum voltage (signal plus common mode) present between either receiver input and receiver signal ground shall not exceed 10 volts nor cause the receiver to malfunction. The receiver shall tolerate a maximum differential signal of 12 volts applied across its input terminals without being damaged.

In the presence of the combinations of input voltages  $V_{ii}$  and  $V_{ib}$  specified in Figure 6/V.11, the receiver shall maintain the specified output binary state and shall not be damaged.

Note. – Designers of terminal equipment should be aware that slow signal transitions with noise present may give rise to instability or oscillatory conditions in the receiving device; therefore, appropriate techniques should be implemented to prevent such behaviour. For example, adequate hysteresis may be incorporated into the receiver to prevent such conditions.



Applied	voltages	Resulting	Output	Purpose of measurement	
Via	VIb	input voltage Vj	binary state		
-12 V 0 V +12 V 0 V	0 V -12 V 0 V +12 V	-12 V +12 V +12 V -12 V	(nat specified)	To ensure no damage to receiver inputs	
+10 V + 4 V -10 V - 4 V	+ 4 V +10 V - 4 V -10 V	+ 6 V - 6 V - 6 V + 6 V	0 1 1 0	To guarantee correct operation at $V_i = 8 \text{ V}$ (maintain correct logic state)	
				300 mV threshold measurement	
+0.30 V 0 V	0 V +0.30 V	+0.3 V -0.3 V	0	} V <sub>cm</sub> = 0 V	
+7.15 V +6.85 V	+6.85 V +7.15 V	+0.3 V -0.3 V	0 1	\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	
-7.15 V -6.85 V	-6.85 V -7.15 V	_0.3 V +0.3 V	1 0	\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	

FIGURE 6/V.11 Receiver input sensitivity measurement

#### 6.4 Input balance test (Figure 7/V.11)

The balance of the receiver input resistance and bias voltages shall be such that the receiver shall remain in the intended binary state under the conditions shown in Figure 7/V.11 and described as follows:

- a) with  $V_i = +720$  millivolts and  $V_{cm}$  varied between -7 and +7 volts;
- b) with  $V_r = -720$  millivolts and  $V_{cm}$  varied between -7 and +7 volts;
- c) with  $V_r = +300$  millivolts and  $V_{cm}$  a 1.5 volt peak-to-peak square wave at the highest applicable signalling rate. (This condition is provisional and subject to further study.);
- d) with  $V_i = -300$  millivolts and  $V_{cm}$  a 1.5 volt peak-to-peak square wave at the highest applicable signalling rate (This condition is provisional and subject to further study).

Note. - The values of V, are provisional and are the subject of further study.

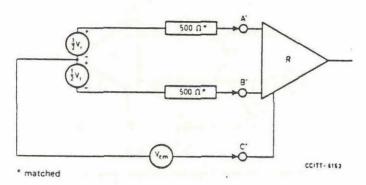


FIGURE 7/V.11 - Receiver input balance test

#### 6.5 Terminator

The use of a terminator is optional depending upon the specific environment in which the interchange circuit is employed (see Annex 1). In no case shall the total load resistance be less than 100 ohms.

#### 7. Environmental constraints

In order to operate a balanced interchange circuit at signalling rates ranging between 0 and 10 Mbit/s, the following conditions apply:

- 1) For each interchange circuit a balanced interconnecting pair is required.
- 2) Each interchange circuit must be appropriately terminated (see Annex 1).
- 3) The total common-mode voltage at the receiver must be less than 7 volts peak. This value is provisional and is subject to further study.

The common mode voltage at the receiver is the worst case combination of:

- a) generator-receiver ground-potential difference (V<sub>e</sub>, Figure 2/V.11);
- b) longitudinally induced random noise voltage measured between the receiver points A' or B' and C' with the generator ends of the cable A, B and C joined together; and
- c) generator bias voltage, if any.

Unless the generator is of a type which generates no bias voltage, the sum of a) and b) above, which is the element of the common mode voltage due to the environment of the interchange circuit, must be less than 4 volts peak.

#### 8. Circuit protection

Balanced generator and load devices complying with this Recommendation shall not be damaged under the following conditions:

- 1) generator open circuit;
- 2) short-circuit between the conductors of the interconnecting cable;
- 3) short-circuit between either or both conductors and point C and C'.

The above faults 2) and 3) might cause power dissipation in the interchange circuit devices to approach the maximum power dissipation that may be tolerable by a typical integrated circuit (IC) package. The user is therefore cautioned that where multiple generators and receivers are implemented in a single IC package, only one such fault per package might be tolerable at any one time without damage occurring.

The user is also cautioned that the generator and receiver devices complying with this Recommendation might be damaged by spurious voltages applied between their input or output points and points C and C' (Figure 2/V.11). In those applications where the interconnecting cable may be inadvertently connected to other circuits, or where it may be exposed to a severe electromagnetic environment, protection should be employed.

#### 9. Detection of generator power-off or circuit failure

Certain applications require detection of various fault conditions in the interchange circuits, e.g.:

- 1) generator in power-off condition;
- 2) receiver not interconnected with a generator;
- 3) open-circuited interconnecting cable;
- 4) short-circuited interconnecting cable;
- 5) input signal to the load remaining within the transition region (± 300 millivolts) for an abnormal period of time.

When detection of one or more fault conditions is required by specific applications, additional provisions are required in the load and the following items must be determined:

- a) which interchange circuits require fault detection;
- b) what faults must be detected;
- c) what action must be taken when a fault is detected, e.g. which binary state must the receiver assume?

The method of detection of fault conditions is application-dependent and is therefore not further specified.

#### ANNEX I

(to Recommendation V.11)

#### Cable and terminations

No electrical characteristics of the interconnecting cable are specified in this Recommendation. Guidance is given herein concerning operational constraints imposed by the length, balance and terminating resistance of the cable.

#### 1. Cable

Over the length of the cable, the two conductors should have essentially the same values of:

- 1) capacitance to ground;
- 2) longitudinal resistance and inductance;
- 3) coupling to adjacent cables and circuits.

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#### 2. Cable length

The maximum permissible length of cable separating the generator and the load in a point-to-point application is a function of the data signalling rate. It is further influenced by the tolerable signal distortion and the environmental constraints such as ground potential difference and longitudinal noise. Increasing the distance between generator and load might increase the exposure to ground potential difference.

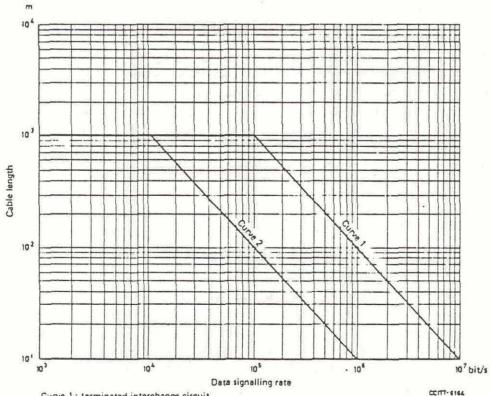
As an illustration of the above conditions, the curves of cable length versus data signalling rate in Figure 8/V.11 may be used for guidance.

These curves are based upon empirical data using twisted pair telephone cable (0.51 mm wire diameter) both unterminated and terminated in a 100 ohm resistive load. The cable length restrictions shown by the curves are based upon the following assumed signal quality requirements at the load:

- 1) signal rise and fall time equal to, or less than, one-half the duration of the signal element;
- 2) a maximum voltage loss between generator and load of 6 dB.

At the higher signalling rates (see Figure 8/V.11) the sloping portion of the curves shows the cable length limitation established by the assumed signal rise and fall time requirements. The cable length has been arbitrarily limited to 1000 metres by the assumed maximum allowable loss of 6 dB.

These curves assume that the environmental limits specified in this Recommendation have been achieved. At the higher signalling rates these conditions are more difficult to attain due to cable imperfections and common-mode noise. Operation within the signalling rate and distance bounds of Figure 8/V.11 will usually ensure that distortion of the signal appearing at the receiver input will be acceptable. Many applications, however, can tolerate much greater levels of signal distortion and in these cases correspondingly greater cable lengths may be employed.



Curve 1: terminated interchange circuit
Curve 2: unterminated interchange circuit

1-IGURT 8/V.11 - Data signalling rate vs cable length for balanced interchange circuit

Experience has shown that in many practical cases the operating distance at lower signalling rates may extend to several kilometres.

For synchronous transmission where the data and signal element timing are transmitted in opposite directions, the phase relationship between the two may need to be adjusted to ensure conformity with the relevant requirements of signal quality at the interchange point.

#### 3. Cable termination

The use of a cable termination is optional and dependent on the specific application. At the higher signalling rates (above 200 kbit/s) or at any signalling rate where the cable propagation delay is of the order of half the signal element duration a termination should be used to preserve the signal rise time and minimize reflections. The terminating impedance should match as closely as possible the cable characteristic impedance in the signal spectrum.

Generally, a resistance in the range of 100 to 150 ohms will be satisfactory, the higher values leading to lower power dissipation.

At the lower signalling rates, where distortion and rise-time are not critical, it may be desirable to omit the termination in order to minimize power dissipation in the generator.

#### ANNEX 2

(to Recommendation V.11)

#### Compatibility with other interfaces

 Compatibility of Recommendation V.10 and Recommendation V.11 interchange circuits in the same interface

The electrical characteristics of Recommendation V.11 are designed to allow the use of unbalanced (see Recommendation V.10) and balanced circuits within the same interface. For example, the balanced circuits may be used for data and timing whilst the unbalanced circuits may be used for associated control circuit functions.

#### 2. Recommendation V.11 interworking with Recommendation V.10

The basic differential receiver specifications of Recommendations V.10 and V.11 are electrically identical. It is therefore possible to interconnect an equipment using Recommendation V.10 receivers and generators on one side of the interface with an equipment using Recommendation V.11 generators and receivers on the other side of the interface. Such interconnection would result in the interchange circuits according to Recommendation V.11 in one direction and interchange circuits according to Recommendation V.10 in the other direction. Where such interworking is contemplated, the following technical considerations must be taken into account:

- 2.1 Interconnecting cable lengths are limited by performance of the circuits working to the Recommendation V.10 side of the interface.
- 2.2 The optional cable termination resistance  $(Z_t)$ , if implemented, in the equipment using Recommendation V.11 must be removed.

#### 3. Recommendation V.11 interworking with Recommendation V.35

Equipment having interchange circuits according to Recommendation V.11 is not intended for interworking with equipment having interchange circuits according to electrical characteristics of Recommendation V.35.

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#### ANNEX 3

(to Recommendation V.11)

#### Multipoint operation

It is considered that further study is required before parameters for this application can be precisely defined and this Annex, giving provisional figures, is intended as a guideline for this study.

#### 1. General

The point-to-point interchange circuit arrangement of one generator and one load might be expanded to a multipoint arrangement by adding generators, receivers or both, at interchange points along the interconnecting cable, as shown in Figures 9/V.11 and 10/V.11.

Only one generator at a given time would present its differential voltage at its interchange point. All other generators would be isolated by an appropriate control, and assume the high impedance state defined below. All receivers would be continuously in an operating condition.

In multipoint configuration, one or more terminators might be required at the cable end or at interchange points, depending on the application. The combined load impedance presented to any active generator by other generators, receivers, cable and terminators, must not be less than 100 ohms.

The operation of a multipoint arrangement must not be perturbed by any of its components when they are either in a high impedance state or a power-off state <sup>61</sup>. The generators and receivers must tolerate without damage the transmitted signals with their maximum amplitude within the specified limits.

Generators on the same multipoint line must have the same nominal d.c. offset voltage in order to operate correctly. However, generators with different d.c. offsets could be used on the same line provided that these differences be compensated at the common reference point.

#### 2. Configurations

Several topological arrangements are to be considered:

- cluster of circuits at the end of a line;
- multidrop line;
- star configuration.

Figure 9/V.11 illustrates a cluster configuration. Each line should be correctly terminated at the receiving end in order to avoid reflections and leads from that point to the receivers kept as short as possible.

Figure 10/V.11 illustrates a multidrop line. The presence of several generators along the line imposes the need to terminate the line correctly at both ends to avoid reflections. The length of the tapping connections along the line must be short enough to avoid mismatching of the main line. A length corresponding to a propagation delay of 1% of the unit interval of the signals seems to be an acceptable limit for the length of the tapping connection.

This tolerable propagation delay and other configurations are to be studied further.

The guidelines on cable length given in Figure 8/V.11 of Annex 1, assuming the same environmental conditions, apply equally to multipoint arrangements. Above 1 or 2 Mbit/s, however, the environmental conditions may be more difficult to control.

<sup>6)</sup> In the power-off state of any device it is assumed that the supply collapses to zero and is replaced by a very low impedance or short circuit.

#### 3. High impedance state

#### 3.1 Static measurements

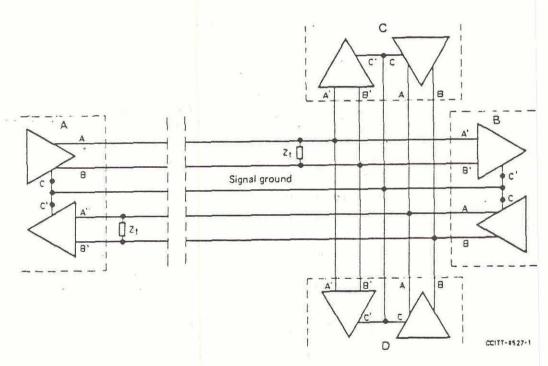
When in the high impedance state and with test loads of 50 ohms connected between each generator output point and point C, the magnitude of the voltage  $V_h$  measured between points A and B shall not exceed 4 mV whatever the logical condition of the generator input data lead (Figure 11/V.11).

When the generator is in the high impedance state, with voltages ranging between -6 V and +6 V applied between each output point and point C, as indicated in Figure 12/V.11, the magnitude of the output leakage currents  $I_{tg}$  and  $I_{th}$  shall not exceed 150  $\mu$ A.

The same condition applies under power-off condition.

#### 3.2 Dynamic measurements

During transitions of the generator output between the low impedance state and the high impedance state, the differential signal measured across a 100 ohm test load connected between the generator points A and B shall be such that the change in amplitude goes from 10% to 90% of the steady state voltage in less than 10µs.



HGURE 9/V.11 Clustered multipoint configuration

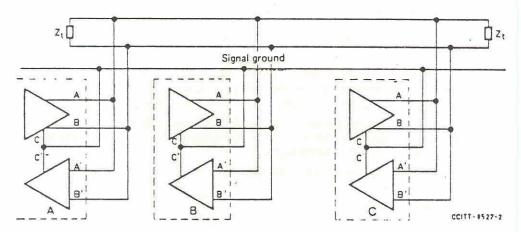
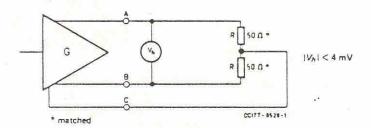
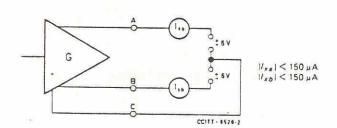


FIGURE 10/V.11 Distributed multipoint configuration



1-1GURI: 11/V.11 High impedance state static measurement



HGURI 12/V.11 Generator output leakage current measurement

APPENDIX F

CONTENT OF DEFINED ND NUMBERS

ND-156 32K RAM	ND-147,190 ADD DUAL PORT	ADD SINGLE	ND-158 ND-145 128K/32 BIT ADD SINGLE BANK POR	ND-146 DUAL BANKS	ND-144 256K BANK	ND-143 128K	S	CONTENT OF	
1132							0- 32K		32
1132									31
1132									30
1132							92-128K	STORAGE	29
			1144	1144		1144		CONTROL	28
		1142	1142	1142		1142	A		27
	1142	1142					В		26
	1142	1142					၁		25
	1142	1142					PORT D	ADDRESS	24
		1143	1143	1143		1143	A		23
	1.143	1143					В		22
	1143	1143					0		21
	1143	1143					PORT D	DATA	20
						1157	Y		1.9
			1145	1145	1145		X-X	LOG	18
	1143	1143					Q		17
	1143	1143					ပ		16
	1143	1143					В		15
			1143	1143	1143		PORT A	DATA	14
	1142	1142					D		13
	1142	1142					C		12
	1142	1142					В		1.1
			1142	1142	1142		PORT A	ADDRESS	0
			1144	1144	1144			CONTROL	6
1132							224-256K		$\infty$
1132							192-224K		1
1132							160-192K		9
1132							128-160K		2
1132							96-128K		4
1132							64- 96K		3
1132							32- 64K		N
1132						24	0- 32K	STORAGE	
			6017	6017	601.7	6017		BACKWIRING	
			YES	YES			WIREWRAP	CRATE+	

ND-06.007.01

BANK Y

### \* \* SEND US YOUR COMMENTS!!! \* \* \* \* \* \* \* \* \* \* Are you frustrated because of unclear information in this manual? Do you have trouble finding things? Why don't you join the Reader's Club and send us a note? You will receive a membership card - and an answer to your comments. Please let us know if you \* find errors \* cannot understand information \* cannot find information \* find needless information Do you think we could improve the manual by rearranging the contents? You could also tell us if you like the manual!! \* \* \* \* \* HELP YOURSELF BY HELPING US!! Manual name: BIG MULTIPORT MEMORY SYSTEM Manual number: ND-06.007.01 What problems do you have? (use extra pages if needed) Do you have suggestions for improving this manual?

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Your name: \_\_\_\_\_ Date: \_\_\_\_\_ Date: \_\_\_\_\_

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Answered by	0		Date	

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